## UNCLASSIFIED

# AD 269 073

Reproduced by the

ARMED SERVICES TECHNICAL INFORMATION AGENCY
ARLINGTON HALL STATION
ARLINGTON 12, VIRGINIA



UNCLASSIFIED

NOTICE: When government or other drawings, specifications or other data are used for any purpose other than in connection with a definitely related government procurement operation, the U. S. Government thereby incurs no responsibility, nor any obligation whatsoever; and the fact that the Government may have formulated, furnished, or in any way supplied the said drawings, specifications, or other data is not to be regarded by implication or otherwise as in any manner licensing the holder or any other person or corporation, or conveying any rights or pensission to manufacture, use or sell any patented invention that may in any way be related thereto.

20692

SOME USEFUL INFORMATION
FOR THE DESIGN
OF AIR-CORE SOLENOIDS

by

D. Bruce Montgomery
J. Terrell

November, 1961 (2nd Printing, December 1961)

Sponsored by the Solid State Sciences Division, Air Force Office of Scientific Research (OAR).

Air Force Contract AF 19 (604)-7344

Principal Investigators: Benjamin Lax and Francis Bitter

National Magnet Laboratory

Massachusetts Institute of Technology

Cambridge 39, Massachusetts

# FOR THE DESIGN OF AIR-CORE SOLENOIDS

PART I: Relationships between Magnetic Field, Power,

Ampere-Turns and Current Density

PART II: Homogenous Magnetic Fields

by

D. Bruce Montgomery
J. Terrell

November, 1961 (2nd Printing, December 1961)

Sponsored by the Solid State Sciences Division, Air Force Office of Scientific Research (OAR).

Air Force Contract AF 19(604)-7344

Principal Investigators: Benjamin Lax and Francis Bitter

Mational Magnet Laboratory

Massachusetts Institute of Technology

Cambridge 39, Massachusetts

#### ACKNOWLEDGEMENT

THE AUTHORS ARE INDEBTED TO F. BITTER FOR HIS MANY CONTRIBUTIONS TO THIS PAPER.

#### **ABSTRACT**

Relationships relating power, magnetic field, current density, and ampere-turns in terms of certain dimensionless factors are summarized for many types of coil geometries and current distributions. A number of plots of these factors are presented.

The field homogeneity in magnet structures is presented in terms of a series expansion about the origin utilizing Legendre Polynomials. A number of tables to facilitate design of homogeneous fields are presented. A method of achieving homogeneity in long solenoid structures by the use of determinants is discussed. Expressions for the axial field from uniform and radially varying current density coils are given.

#### PART I

### RELATIONSHIPS BETWEEN MAGNETIC FIELD, POWER, AMPERE-TURNS AND CURRENT DENSITY

It has been customary in the literature on magnet design to write the relationships between fields, power, ampere-turns and current density in terms of a number of dimensionless factors. In this section we will treat in summary form twelve cases of coil construction (Cases I - XII). This work is in part a summary, and enlargement of parts of the following three papers: F. Bitter, R.S.I. Vol. 7, 1936 Part II; W. F. Gauster, AIEE, Fall General Meeting, Chicago, October 1959; F. Gaume, Journal de Recherche du Centre Nationale de Recherche Scientifique, No. 43, June 1958. (English translation, Lincoln Laboratory, M81-12, part I and II, Dr. H. H. Kolm.)

A. The most commonly used dimensionless factor is the G factor first suggested by Fabry, which connects a given magnetic field with the power required to produce this field and is a function of the type of current distribution and the geometry of the coil.

$$H = G \left(\frac{W\lambda}{\rho a_1}\right)^{1/2} \tag{1}$$

with

W = watts

 $\rho$  = resistivity in ohm - cm

Q1 = inner radius in cm

 $\lambda$  = fractional volume of conductor

The space factor  $\lambda$  is assumed constant throughout the volume. If  $\lambda$  is a function of x and y, were complicated formulas are needed.

B. A second dimensionless factor J has been suggested by W. F. Gauster and connects the current density at the innermost winding of a magnet to the power of the magnet. J is also a function of the current distribution and the geometry of the coil.

$$j_1 = J \left( \frac{W}{\rho \lambda \alpha_1^3} \right)^{1/2} \tag{2}$$

C. rothin, inticular type of coil a maximum G factor can be found which will give the most magnetic field for the least power used. However, in many types of coil construction this maximizing of the G factor may lead to excessively high current densities at the innermost winding which may make cooling the coil difficult. To illustrate this point a ratio,  $\gamma$ , can be formed which compares the maximum current density in any coil to the current density in a uniform current density coil, where both have the same inner radius and both develop the same field. Quantities for the uniform case are designated as primed:

$$H = G \left(\frac{W\lambda}{\rho a_1}\right)^{1/2} \quad H' = G' \left(\frac{W'\lambda}{\rho a_1}\right)^{1/2}$$

$$if \quad H = H'$$

$$\left(\frac{W}{W'}\right)^{1/2} = \frac{G'}{G}$$

$$\frac{j_1}{j_1'} = \frac{J}{J'} \left(\frac{W}{W'}\right)^{1/2}$$
and using (3)
$$\frac{j_1}{j_1'} = \frac{J}{J'} \frac{G'}{G} = \gamma$$
(4)

This ratio indicates the difficulty of cooling and supporting any coil compared with a coil of uniform current density at the same field and bore.

D. The ratio in equation (3) is also useful and represents the power saving in using any current distribution compared with uniform current density when generating the same field in the same bore. We give it in the more useful form.

$$\frac{W}{W'} = \left(\frac{G'}{G}\right)^2 = P \tag{5}$$

These dimensionless factors and ratios are presented in Table I for nine types of magnet construction and following Table I for three additional more complicated magnet constructions.

One of the most interesting new coil constructions presented is that of case V. This coil was first suggested by F. Gaume and achieves a relatively high G factor without excessive current densities and can be quite simply constructed of plates in the usual Bitter magnet style, with the introduction of axial current variations by making the plates thicker towards the ends of the magnet.

A number of plots of the geometry factor, G, against  $\alpha$  and  $\beta$  are given in Figures 1 through 6. Cockcroft's curves for case VIII are reprinted in Figure 1, Bitter's curves for case VI in Figure 2, new curves for cases VIII and VI in the small  $\alpha$  and  $\beta$  region are given in Figures 3 and 4. New curves for case V (Gaume's axial current variation) and for case X (the ellipsoidal shell) are given in Figures 5 and 6. The variables  $\alpha$  and  $\beta$  are defined:

$$\alpha = \frac{a_2}{a_1}$$

$$\beta = \frac{b}{a_1}$$

E. It is sometimes instructive to write the magnetic field in terms of ampereturns rather than power. This arises from the practical standpoint that it is difficult to always predict the resistance of a magnet exactly, especially as a function of power, and one may wish to know how much field will be produced for each ampere available in the power supply. This relationship between ampere-turns and field can be related to the J factor and the G factor and miscellaneous  $\alpha$  and  $\beta$  terms and numerical constants. The relationships are given here for three cases: Case VI, Case VII and Case VIII. All the others are easily derivable by integrating the current density over the coil to obtain the total current in terms of J and W. Total current I and the factor J can now be substituted for W in equation (1), and the ampere-turns related to the field.

Case VI: 
$$H_{6} = \frac{I\lambda}{t} \left[ \frac{1}{J_{5} \ln \alpha} \right] G_{6}$$

$$= \frac{NI}{2b} \left[ \frac{1}{J_{6} \ln \alpha} \right] G_{6}$$
Case VII: 
$$H_{7} = \frac{I\lambda}{t} \left[ \frac{1}{J_{7}} \right] G_{6}$$

$$= \frac{NI}{kg_{1}} \left[ \frac{1}{(\alpha - 1)J_{7}} \right] G_{7}$$
(6)

Case VIII: 
$$H_{8} = \frac{I\lambda}{t} \left[ \frac{1}{J_{8}(\alpha - 1)} \right] G_{8}$$

$$= \frac{NI}{2b} \left[ \frac{1}{J_{8}(\alpha - 1)} \right] G_{8}$$
(8)

F. Useful geometry factors can be written for very long solenoids (long enough so that end effects can be neglected).

For a constant current density coil the ampere-turns per unit length can be written:

$$\frac{NI}{2b} = \left(\frac{(W/2b)\lambda}{2\pi\rho}\right)^{1/2} \left(\frac{2(\alpha-1)}{\alpha+1}\right)^{1/2}$$
 (9)

where W/2b is the power per unit length (watts/centimeter)
For the disk construction, with a radial current distribution we write

$$\frac{NI}{2b} = \left(\frac{(W/2b) \lambda}{2\pi \rho}\right)^{1/2} \left(\ln \alpha\right)^{1/2} \tag{10}$$

Rewriting (9) and (10) in terms of the field in gauss and power per length as watts per meter, we obtain

$$H(gauss) = \frac{(2\pi)^{1/2}}{50} G_5 \left( \frac{W(\text{watts/meter}) \lambda}{P(\text{ohm cm.})} \right)^{1/2}$$

where

$$G_5$$
 (uniform) =  $\left(\frac{2(\alpha-1)}{\alpha+1}\right)^{1/2}$ 

$$G_5(\text{radial}) = (\ln \alpha)^{1/2}$$

These factors are plotted in Figure 7 and it is interesting to note that below an Q of 2.5 there is negligible power saving with the disks, but at greater Q, the saving is increasingly pronounced; at an Q of 10, 45% more power would be required by the uniform wound solenoid for the same magnetic field.

G. It is useful to use the G factor notation when dealing with split coils (coils separated by a gap). If we take a pair of coils with parameters as defined in Figure 8, we can write the field in the gap by simple addition and subtraction of coils.

$$\alpha = \frac{a_2/a_1}{2a_1}$$

$$\beta_1 = \frac{\text{total length}}{2a_1} = \frac{4b + lg}{2a_1}$$

$$\beta_2 = \frac{gap}{2a_1} = \frac{lg}{2a_1}$$

Figure 8

$$H(0,0) = \left(\frac{W\lambda}{\rho a_1}\right)^{1/2} \left(\frac{G_1 \beta_1^{1/2} - G_2 \beta_2^{1/2}}{(\beta_1 - \beta_2)^{1/2}}\right)$$

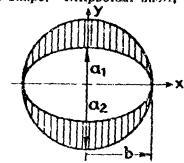
$$G_1 = G_1(\alpha_1 \beta_1)$$
,  $G_2 = G_2(\alpha_1 \beta_2)$ 

This relationship (11) holds for any current distribution having no axial variation.

TABLE 1

			=		Square ends	H6 라스 스		APERCON FERENCE	
Power Ratio for Some Field Referred to Cose VIII	0 421	0.495	0.565	1990	0.592	921.0	C.78	1.000	1.065
Current Density Ratio for Same Field Referred to Case XIII	2.74	2.75	¥2.3	3.83	1.40	1.30	<b>3</b> 5.1	1:00	1.17
Jack.	a - 10 a275 0.421	G-B-10 G-255 0.392	k-1, a-10 0138 0.298	6-10, 6-3 G216 O.463	G. 232 G. 232 O.181	9.209 0.209 0.149	k-1, 0.45 G201 Q.151	a-5, a-2 G179 a099	k=1, d=2.7 G=.172 3.135
J Current Denaity Factor	( <u>a</u> ) ( <u>a</u> )	- 100	((3-1) 4 Ph.) A	-18	0-1 106 a lang	*( 1 1 1 ) *	( 1 ( 1 ( 1 ( 1 ( 1 ( 1 ( 1 ( 1 ( 1 ( 1	(200(04.1))	(40k(01))W
Gmax.	G = 145. G 289	e, å = ine 0.27£	h=1, G= inc 0.250	G - mt , A - B 0.125	F . A - 6 0.232	0200	k-1, C=45 0201	G=3, D=E 0.179	k.1, e.£7 0.17£
G Geometry Factor	$\frac{1}{3} \left( \frac{2\pi}{2} \right)^{1/2} \left( \frac{d-1}{d} \right)^{1/2}$	$\frac{(2\pi)^{16}}{10} \left( \frac{1}{4} \left( \frac{1}{1.67}, \frac{\Delta}{G^{+}_{1,1}} \right) \right)^{12} \\ - \frac{\Delta}{4\pi} \ln^{1} \frac{\Delta}{G} + \frac{\Delta}{4} \ln^{1} \frac{\Delta}{G} \right)^{12}$	\$ (15.7) (2) ** (Max when k=1)	10 (2 to to 2) 01	(1) (1) (1) (1) (1) (1) (1) (1) (1) (1)	(a)	1 ( 2k) W L C 5 (2k1) ( (2l.))72 (Mox when K-1)	(1 ) 1 1 0 (5 ( ) ) 1 1 ( ( ) ) 1 1 1 1 1 1 1 1 1 1 1	1 (3pk) <sup>R</sup> (g-1) 5 (\$1*1) (g-1)V (Max when X-1 )
Current Distribution	Optimum Current Distribution $1 - f(x,y) \sim \frac{y}{x(x^2 + y)}$	Optimum Current Datribrion 1 := f(e,y) ~ (ee,y) > 4	Detumum Current Distribution 1 to 17 For Tapered Ends	Correct Describution tof(y) ~ V(6*+y*)	1 ~ 1 [ ( ( ( ( ( ( ( ( ( ( ( ( ( ( ( ( ( (	- <del>1</del> ~ i	-1	( - Content	l = Caration
Shape of Coli		Square Engl	Spends of Scott	Se see fran	Cyndric Bus Score Ends	Cynneruc' Bory Square Erds	Cylindrice Bon Terres Ends	Space End	Cymerical Box
•	-	Ħ	Н	Ħ	SCURE S	12 S	耳		Ħ

Coil Shape: Ellipsoidal sholl, & = constant



$$\alpha = \frac{\alpha_2}{\alpha_1}$$

$$\beta = \frac{h_{max}}{a_1}$$

Current Distribution: i = io

Ceometry Factor:

i) 
$$\frac{\beta > 1}{\beta > \alpha}$$
  $G = \frac{\pi^{1/2}}{5 \ln \alpha} \beta^{1/2} \left( \frac{1}{(\beta^2 - 1)^{1/2}} \ln (\beta + (\beta^2 - 1)^{1/2}) - \frac{1}{(\beta^2 - \alpha^2)^{1/2}} \ln \left( \frac{\beta + (\beta^2 - \alpha^2)^{1/2}}{\alpha} \right) \right)$ 

ii) 
$$\frac{\beta < 1}{\alpha > \beta}$$
  $C = \frac{\pi^{1/2}}{5} \beta^{1/2} \left( \frac{1}{(\beta^2 - 1)^{1/2}} \sec^{-1} \frac{1}{\beta} - \frac{1}{(\alpha^2 - \beta^2)^{1/2}} \sec^{-1} \frac{\alpha}{\beta} \right)$ 

iii) 
$$\frac{\beta > 1}{\alpha > \beta}$$
  $G = \frac{\pi^{1/2}}{5} \beta^{1/2} \left( \frac{1}{(\beta^2 - 1)^{1/2}} \ln (\beta^2 - 1)^{1/2} \right)$ 

$$-\frac{1}{(\alpha^2-\beta^2)^{1/2}} \sec^{-1}\frac{\alpha}{\beta}$$

 $G_{\text{max}} = .204 \text{ at } \alpha = 6, \beta = 2$ 

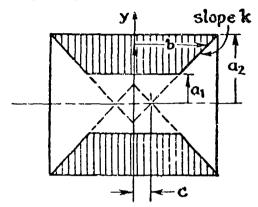
Current Density Factor J:

$$J = \left(\frac{1}{4\pi\beta \ln a}\right)^{1/2}$$
at  $a = 6$ ,  $\beta = 2$ ,  $G_{max} = 0.204$   
and  $J_{max} = 0.149$ 

Current Density Ratio Y for same Field referred to case VIII: = 1.31
Powe: Ratio P for same Field referred to case VIII = 0.769

#### CASE XI

Coil Shape: Cylindrical Bore, tapered ends, non-coinciding apexes.



$$b = ky + c$$

$$k = slope$$

$$y = \frac{c}{a_1} \quad \alpha = \frac{a_2}{a_1}$$

Current Distribution:

Geometry Factor:

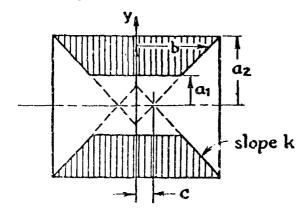
$$G = \frac{\pi^{1/2}}{5} \frac{1}{k(\alpha-1)+y \ln \alpha} \left[ \frac{k}{k^2+1} \cdot \frac{1}{k(\alpha-1)+y \ln \alpha} \left[ \frac{k}{k^2+1} \cdot \frac{(k^2+1)\alpha^2+2ky\alpha+y^2)^{1/2}+\alpha(k^2+1)+y(k^2/k^2+1)^{1/2}}{((k^2+1)+2ky+y^2)^{1/2}+(k^2+1)^{1/2}+y(k^2/k^2+1)^{1/2}} + \ln \left( \alpha \frac{((k^2+1)+2ky+y^2)^{1/2}+y+k}{((k^2+1)\alpha^2+2ky\alpha+y^2)^{1/2}+y+k\alpha} \right) \right]$$

G is a maximum when k = 1 and C = 4.5 which reduces case XI to case VII Current Density Factor J:

$$J = \left(\frac{1}{4\pi(k(\alpha-1)+\gamma \ln \alpha)}\right)^{1/2}$$

#### CASE XII

Coil Shape: Cylindrical Bore, tapered ends, non-coinciding apexes.



$$b = ky + c$$

$$k = slope$$

$$y = \frac{c}{a_1} \quad \alpha = \frac{a_2}{a_1}$$

Current Distribution:

$$i = \frac{i_0}{b} = \frac{i_0}{ky+c}$$

Geometry Factor:

$$G = \frac{\pi^{1/2}}{5} \frac{k}{(k^2+1)^{1/2}} \left[ \left( \frac{1}{k(\alpha-1) + \gamma \ln(\frac{\gamma+k}{\gamma+\alpha k})} \right) \cdot \int_{m} \frac{((k^2+1)\alpha^2 + 2k\gamma\alpha + \gamma^2) + \alpha(k^2+1)^{1/2} + \gamma \left( \frac{k^2}{k^2+1} \right)^{1/2}}{((k^2+1)+2k\gamma+\gamma^2)^{1/2} + (k^2+1)^{1/2} + \gamma \left( \frac{k^2}{k^2+1} \right)^{1/2}} \right]$$

G is a maximum when k = 1.0 and  $\alpha = 4.5$  which reduces case XII to case VII Current Density Factor J:

$$J = \left(\frac{1}{4\pi (k(\alpha-1) + \gamma \ln(\frac{k+\gamma}{k\alpha+\gamma}))}\right)^{1/2}$$

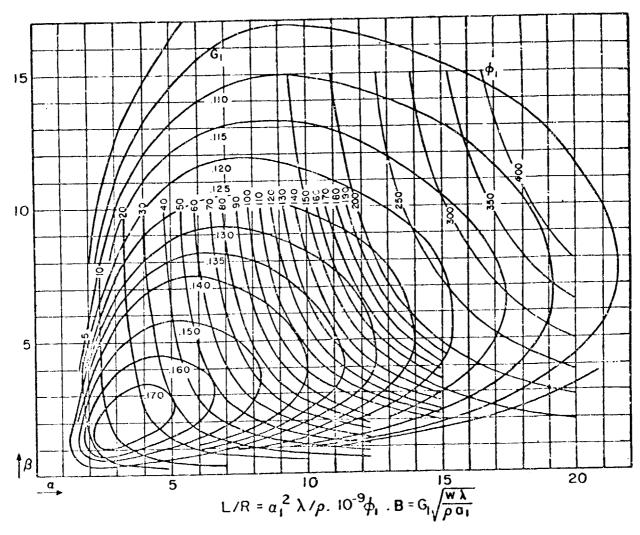
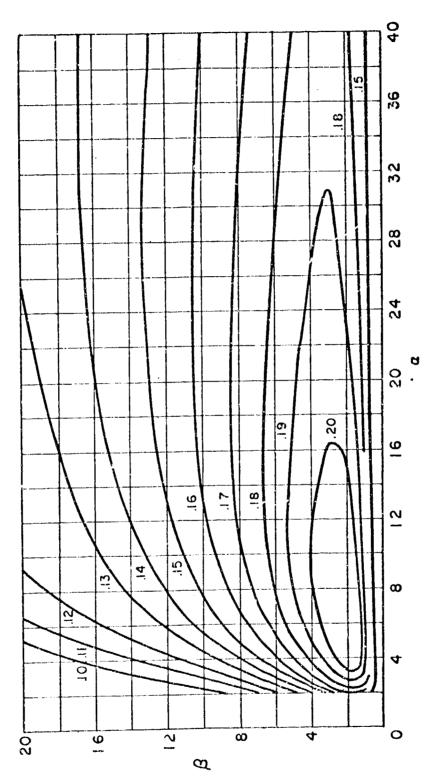


Fig. I Reproduction of Cockcroft's G-factor curves for uniformly wound coils with a square cross-section (Case VIII)



Reproduction of Bitter's G-factor curves for plate magnets, i = io/r (Case VI) Fig. 2

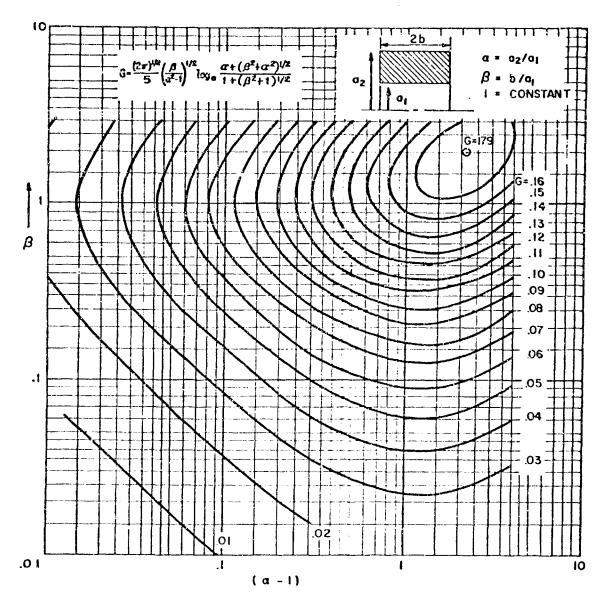


Fig. 3 G-factors for uniform current density magnets with small  $\alpha$  and  $\beta$  's (Case VIII)

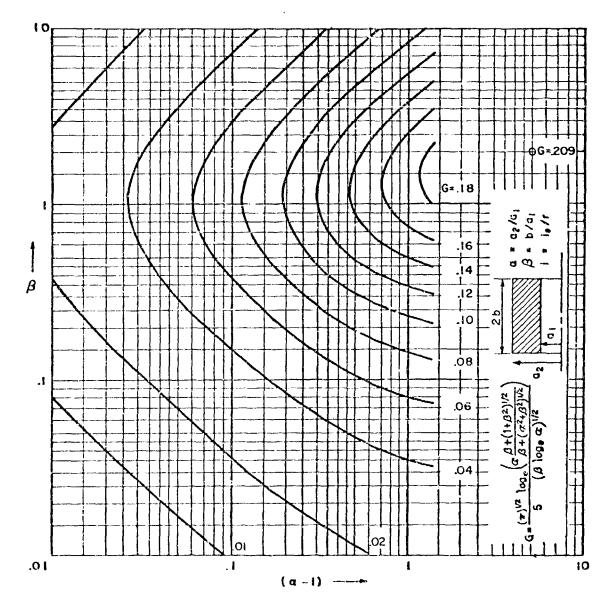


Fig. 4. G-factors for radial current density magnets with small  $oldsymbol{a}$  and  $oldsymbol{eta}$ 's (Case VI)

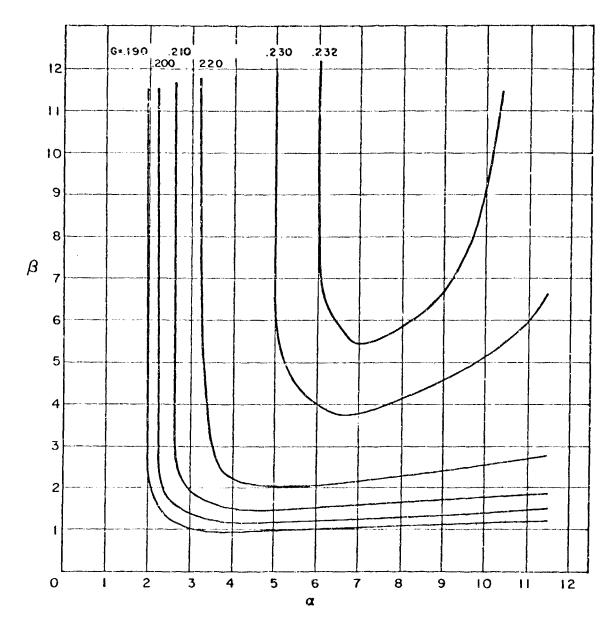
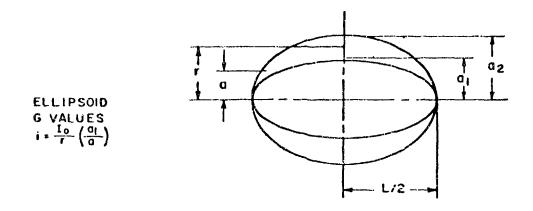


Fig. 5 G-factors for a Gaume current distribution (Case V)



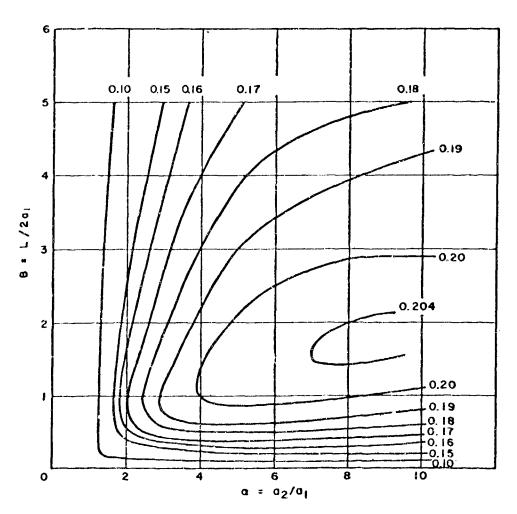


Fig. 6 G-factors for an ellipsoid of constant α (Case X)

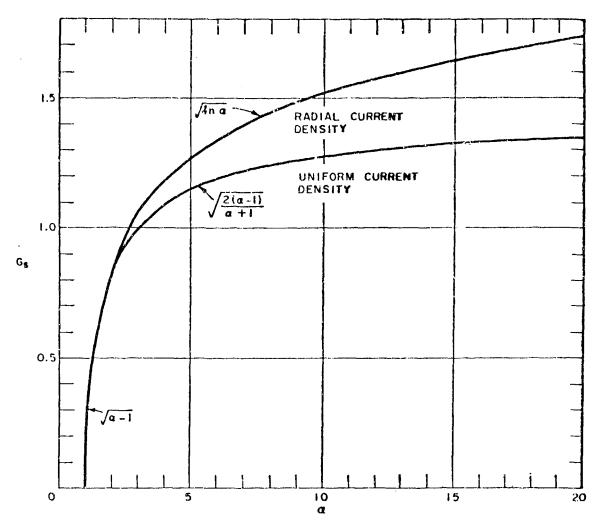


Fig. 7 G-factors for long solenoids

A.

A magnetic field can be written as a power series involving Legendre polynomials expanded about the origin of the field. If this origin is on the axis and on the plane of symmetry the expansion will have no odd terms and the axial component  $H_Z$  and the radial component  $H_D$  can be written:

$$H_{z}(r_{1}\theta) = \sum_{n=1}^{\infty} \frac{1}{(2n-2)!} \left[ H_{z}^{(2n-2)}(z_{1}0) \right] r^{2n-2} p_{2n-2}(u)$$
(1)

$$H_{\rho}(r,\theta) = -\sum_{n=1}^{\infty} \frac{1}{(2n-1)!} \left[ H_{z}^{(2n-2)}(z,o) \right] r^{2n-2} P_{2n-2}'(u)$$
(2)

$$u = \cos \theta$$

In these equations the coefficients are defined as follows:

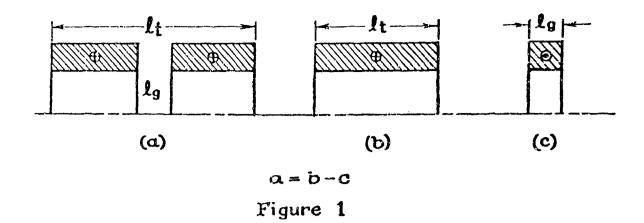
$$H_z^{(2n-2)}(z_10) \equiv \frac{d^{2n-2}H_z(z_10)}{dz^{2n-2}}$$

and P2n-2(u) are the Legendre polynomials tabulated in Table 2.

The importance of this approach, due to W. Garrett<sup>1</sup>, is that the field at any point on the axis of a system of coaxial coils can be given explicitly in simple algebraic form. From this, the derivatives of equations (1) and (2) can be obtained, and so a complete expression for the field in the vicinity of the origin can be obtained. If sufficient derivative terms are used the field can be found anywhere within a sphere whose center lies at the origin and which does not extend far enough to include any corners of the coil. However, out to within a few percent of the inner radius of the coil, the field can be found quite accurately with only a few derivative terms. Compensation can be designed for each component of the field separately. In general, if we have a coil system of the assumed symmetry having N variable parameters such as size, or ampere-turns, it is possible to specify N coefficients in the expansions 1 and 2.

For example, to examine the standard Helmholtz pair of coils, where the coils are so spaced that there is no second derivative of the field at the origin, we can think of the coil system in the following way:

<sup>1.</sup> M. W. Garrett, J. Appl. Phys. 22, 9, Sept. 1951



As shown in Fig. 1, we consider a single coil having length equal to the coils and the gap separation, from which we subtract a second coil of the length of the gap. The gap coil is thought of as having a current in the reverse direction to the original long coil. We now merely need select the length of the gap coil so that it will have a second derivative equal and opposite to that of the original long coil and the two will thus cancel at the crigin.

There are many other combinations which would also cancel the second derivative at the origin, for instance, the hypothetical gap coil with its reverse current could have been somewhat longer and had less reverse current in it, or shorter and had more reverse current in it. Both of these solutions, of course, lead to a continuous solenoid with no gap but with an altered current density in the center section. These various solutions will have both an effect on the amount of field in the center of the magnet and on the size of the higher derivatives. The greater the distance over which one compensates, the more parallel will be the flux lines from the compensating coil, hence the flatter will be the flux distribution at the origin, and hence the smaller will be the sum of the higher derivatives. This compensation may not be compatible with obtaining the highest central field for the least power, however.

B.

Equations (1) and (2) can be rewritten in the following form:

$$H_{Z}(\mathbf{r}_{1}0) = H_{Z}(\mathbf{0}_{1}0) \left[1 + \mathcal{E}_{Z}\left(\frac{\mathbf{r}}{\mathbf{q}_{1}}\right)^{2} P_{Z}(\mathbf{u}) + \mathcal{E}_{A}\left(\frac{\mathbf{r}}{\mathbf{q}_{1}}\right)^{4} P_{A}(\mathbf{u}) + \cdots\right]$$
(3)

$$H_{P}(\mathbf{r}_{1}0) = H_{Z}(\mathbf{0}_{1}0) \left[0 + \mathcal{E}_{Z}\left(\frac{\mathbf{r}}{\mathbf{q}_{1}}\right)^{2} P_{Z}'(\mathbf{u}) + \mathcal{E}_{A}\left(\frac{\mathbf{r}}{\mathbf{q}_{1}}\right)^{4} P_{A}'(\mathbf{u}) + \cdots\right]$$
(4)

$$H_{Z}(\mathbf{0}_{1}0) = G\left(\frac{W\lambda}{Pa_{1}}\right)^{1/2}$$

2. R. S. Ingarden and J. Michalczyk, Bulletin De L'Academie Polonaise Des Sciences (Serie des sci. math, astr., and phys) VIII, 5, 1960.

The expansion has been normalized to the inside radius of the coils a1. A tabulation of the  $\mathcal{E}_{\mathrm{H}}$  coefficients for three useful geometries is given in Table 1. The geometries treated are (i) a cylindrical thin current sheet, (ii) a finite solenoid with a uniform current density, and (iii) a finite solenoid with a radial current distribution (i=10/r). The value of the first eight Legendre polynomials and their derivatives are given for convenience in Table 2. From equations (3) and (4) the magnetic field can be found at any point near the origin of a magnet, at a distance  $\boldsymbol{r}$  from the origin and at an angle  $\boldsymbol{\theta}$ . The derivative terms in equation (3) have their largest value for  $\theta = 0$  (P<sub>n</sub> (1) = 1 for all n when  $\Theta = 0$  and  $\cos \Theta = 1$ ). The maximum value of the terms therefore can be obtained from the simpler form of (3) where  $\theta = 0$  and z = r:

$$H_{z}(Z_{1}0) = H_{z}(0_{1}0)\left[1 + \mathcal{E}_{2}\left(\frac{z}{\alpha_{1}}\right)^{2} + \mathcal{E}_{4}\left(\frac{z}{\alpha_{1}}\right)^{4} + \cdots\right]$$
 (5)

C.

The most commonly used method of achieving some degree of homogeneity in coils is to use the spaced helmholtz pair. Garrett has compiled a number of tables dealing with this method of compensation and since these tables have not been published we are making them part of this report. These tables are for constant current density only. The first table, Table 3 is used to determine the proper spacing between finite coils to achieve cancellation of the second derivative at the origin. The second table, Table 4 is used to determine how far from the origin one can move before the field deviates by more than 0.1 percent of the field at the origin. The third table, Table 5 is particularly useful and can show how accurately the coils must be spaced in order to cancel the second derivative to the desired degree. The two remaining tables, Table 6 and Table 7 give the resultant fourth and the sixth & coefficients remaining after the proper spacing of the coils. The tables are used with the notation of equations (3) and (4) except that Garrett has normalized the expansion to the mean radius of the coil rather than to the inner radius as in Section B above ( $(\frac{z}{\alpha_0})$  rather than  $(\frac{z}{\alpha_1})$  where  $\alpha_0 = (\frac{\alpha_1 + \alpha_2}{2})$ ). To illustrate the use of Garrett's tables, we work out the following examples:

Assume we have two coils, each with an  $\alpha = 3$  and a  $\beta = 1$  and we wish to separate them to cancel the second order derivative at the origin.

$$X = \frac{4\beta}{\alpha + 1} = 1.0$$

$$A = \frac{2(\alpha - 1)}{\alpha + 1} = 1.0$$

From Table 3, 
$$K = .58200 - \frac{\Delta X}{a_1 + a_2} - \frac{\Delta X}{a_1(1 + \alpha)}$$

$$\frac{\Delta X}{a_1} = 2.33040 = 2\beta + \frac{19}{a_1}$$

and the spacing should be

$$\frac{\text{lg}}{a_1} = .33040$$

Using Table 4 we see that  $V = \frac{Z}{a_0} = .181$  or  $\frac{Z}{a_1(1+\alpha)} = .181$  and  $\frac{Z}{a_1} = .724$ . This means that out to a distance along the axis of 72.4% of the inner radius, the field is within 0.1% of the field at the origin. Using Table 5, we can examine the effect of improper spacing.

$$H = H_0 \left( 1 + Q \left( \frac{d}{a_0} \right) \cdot \left( \frac{z}{a_0} \right)^2 + \mathcal{E}_4 \left( \frac{z}{a_0} \right)^4 + \cdots \right)$$

where Q is the value in table 5 and d/a<sub>0</sub> is the displacement error in locating one coil. For the above example, we can find the error that will make the uncancelled 2nd derivative error equal the fourth at  $(\frac{z}{do}) = 0.1$ 

Q = 4.21 (at X = 1, A = 1)  

$$\mathcal{E}_4$$
 = .922 from table 6  
4.21  $\left(\frac{d}{a_0}\right) \times 10^{-2} = .922 \times 10^{-4}$   
 $\frac{d}{a_0}$  = 2.18 × 10<sup>-3</sup>

D.

When maximum homogeneity and not maximum efficiency is of the greatest importance, the higher derivatives must be handled. Since the number of derivatives that can be cancelled depends upon the number of variables, we must now pick solutions which have more variables at our disposal. For instance, if we wish to cancel both the second and the fourth derivative of the field we can superimpose two coils on each other each with two geometry variables or each with one geometry variable and one current variable. To proceed, then, we would pick the length of one coil and the current of one coil and the variables then would be the length of the second coil and the ratio of the two current densities. Garrett has calculated a set of so called sixth-order solenoids (no second or fourth derivatives) by making the following choice of variables: He constructs an overwound end solenoid by winding N layers of turns over a certain central region of the solenoid and 2N layers over a certain section of both ends (i.e. a notch on the center section of the o.d. of the coil). He now solves uniquely for the length of the overall coil and the length of the notch. The results of his calculation are shown in Table 8 for several thicknesses of coil.

There are other choices of two variables that one can make to construct sixth-order solenoids and Table 9 illustrates several methods. Table 9 was constructed for the specific example of a rather long solenoid. Several general remarks can be drawn from the table. Probably the most important of which is that since the selenoid is very long

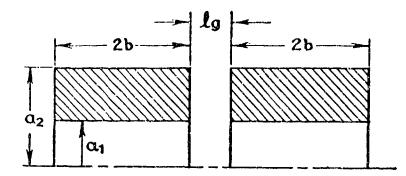
the job of compensating it for high homogeneities is much simplified, the higher derivative having been quite small to begin with. We notice, however, that in all but one case in the table, cancelling the second and fourth derivatives of the field, have increased the sixth derivative. The single exception is the current sheet of the o.d. The current sheet on the o.d., however, requires more power than any of the other methods of solution. Perhaps the most attractive method of solution is that represented by the last column, that of current sheets on either end of the i.d. of the coil. This method takes little power and leaves the center section of the magnet open for further compensation if desired.

Cancellation of higher derivatives than the fourth is of course possible, but is often of decreasing practical importance because of the unrealistic accuracy necessary in the cancellation of lower derivatives, if the leftover uncompensated part of say the second, is to be smaller than the sixth derivative or eighth derivative, etc. It probably makes sense to design coils to cancel the second and the fourth derivative and use a small auxiliary set of coils driven from the separate power source to tune out any random errors which creep in. Some of these random errors are of course going to be nonsymmetrical and will give rise to odd derivative terms. It is quite possible to construct a set of auxiliary compensating coils which will operate independently of each other. One coil can be designed to have a second derivative and no fourth and a second coil to have a fourth derivative and no second, etc. Each one can then be tuned independently of the other.

E.

We have constructed a table of  $\mathcal{E}$  coefficients up to the eighth order for coils of varying  $\alpha$  and  $\beta$  with both constant current density and with current density equal  $i \approx i \sigma/p$ . The tables also include the G factor and the multiplication of the G factor and the  $\mathcal{E}$  coefficients. The tables are to be used with the expansion of equation (3) and (4) where z is referred to the inner radius  $\alpha_1$ ;  $\mathcal{E}$  is tabulated as D for uniform currents and as A for the radial current distribution. Table 10A gives the coefficients for the uniform current density case for several  $\alpha$  and  $\beta$  up to 10. Table 10B gives the coefficients over the same range of  $\alpha$  and  $\beta$  for the radial current distribution case. Tables 11A, B, C, D show more detail for the constant current and the radial current cases in the small  $\beta$  range.

The use of the tables is probably best illustrated by the following example: Let us assume we have a coil we wish to separate for purposes of access and we want to know the resultant field in the gap and the resultant homogeneity (see Fig. 2)



W = power  $\lambda = space factor$   $\rho = resistivity$ 

$$\alpha = \frac{\alpha_2}{\alpha_1}$$

$$\beta_t = \frac{4b + lg}{2a_1}$$

$$\beta_g = \frac{lg}{2a_1}$$

$$G_g \left(\alpha_1 \beta_g\right)$$

Figure 2

If we now subtract from the field expansions for a coil of length  $4b+\mbox{$l$}_{g}$ , a coil of length  $\mbox{$l$}_{g}$  we can write:

$$\begin{split} H_{gap} &= \left(H_{0t} - H_{0g}\right) + \left(H_{2t} - H_{2g}\right) \left(\frac{z}{a_{1}}\right)^{2} + \left(H_{4t} - H_{4g}\right) \left(\frac{z}{a_{1}}\right)^{4} + \cdots \\ H_{gap} &= \left(\frac{W\lambda}{\beta a_{1}}\right)^{1/2} \left(\frac{\beta_{t}}{\beta_{t} - \beta_{g}}\right)^{1/2} \left[\left(G_{t} - \frac{G_{g}\beta g^{1/2}}{\beta t^{1/2}}\right) + \left(G_{t} \mathcal{E}_{2} - \frac{G_{g}\beta g^{1/2}}{\beta t^{1/2}} \mathcal{E}_{2}^{i}\right) \left(\frac{z}{a_{1}}\right)^{4} + \cdots \right] \\ &= \left(G_{t} \mathcal{E}_{4} - \frac{G_{g}\beta g^{1/2}}{\beta t^{1/2}} \mathcal{E}_{4}^{i}\right) \left(\frac{z}{a_{1}}\right)^{4} + \cdots$$

where the primed quantities refer to the error coefficients for the gap. Taking a specific geometry, let the current be constant ( $\mathcal{E} = D$ ).

$$\beta_{t} = 4$$

$$\alpha = 3$$

$$\beta_{g} = 0.2$$

What is the field in the gap and the second error coefficient, and what would be the second error coefficient if there were no access gap?

The  $D_2$  listed for  $\alpha = 3$ ,  $\beta = 4$  in Table 10A is - .1415 x  $10^{-1}$ , therefore the second error coefficient of the whole coll would be  $\mathcal{E}_2 = -1.415 \times 10^{-2}$  where

$$H = H_0 \left(1 + \mathcal{E}_2 \left(\frac{z}{a_1}\right)^2 + \cdots\right)$$

To solve for the split coil, we use (6) and the  $G_1$ .  $D_1$  columns in Table 10A

$$H_{gap} = \left(\frac{W\lambda}{\rho_1 a_1}\right)^{1/2} \left(\frac{4}{3.8}\right)^{1/2} \left[ \left(0.1580 - 0.08639 \left(\frac{0.2}{4}\right)^{1/2}\right) + \left(-0.2235 \times 10^{-2} + 0.5004 \times 10^{-1} \left(\frac{0.2}{4}\right)^{1/2}\right) \left(\frac{z}{a_1}\right)^2 + \cdots \right]$$

$$H_{gap} = \left(\frac{W\lambda}{\rho a_1}\right)^{1/2} \left(1.03\right) \left[ .1387 + 0.8965 \times 10^{-2} \left(\frac{z}{a_1}\right)^2 \right]$$

this gives a central field of

$$H_{gap}(0.0) = \left(\frac{W\lambda}{\rho a_1}\right)^{1/2} (0.143)$$

and if the second order coefficient is written in the following form:

$$H = H_0 \left( 1 + \mathcal{E}_2 \left( \frac{z}{a_1} \right)^2 + \cdots \right)$$

the coefficient for the above example would be

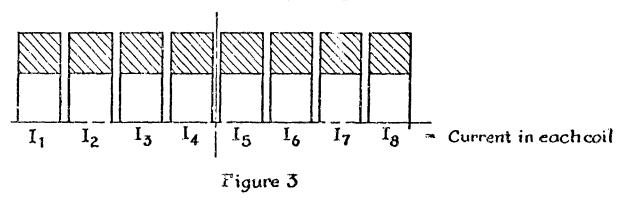
$$\mathcal{E}_2 = \frac{0.8965 \times 10^{-2}}{.1387} = 6.46 \times 10^{-2}$$

The loss in field for this case has been 9.5% and the second order error coefficient has been increased by a factor of 4.55. If we had been trying to homogenize the field by separating the coils, we note that we would have separated them by too much. To use these tables for synthesis (i.e. cancellation of derivatives) it would be necessary to use trial and error, in the manner of the above example.

It is interesting to note the oscillatory behavior of the  $\mathcal{E}$  coefficients, particularly in the small  $\beta$  region. The second derivative is always negative, the fourth derivative goes through zero once, the sixth derivative goes through zero twice and the eighth derivative three times. This means that there are colls which can be built with no fourth, no sixth and no eighth derivatives. This may be useful in the design of independent compensating colls.

If one wishes to compensate a very long solenoid over a considerable length of the axis such as a multicoil structure used for plasma experiments, expansion of the derivatives around the origin has limited usefulness. One might wish to use the following procedure instead; namely specifying that the field at any number of intervals along the axis be constant and then solving for the current distribution in the coils that would make the field be equal at the specified points. In the case of a multicoil structure one could ask that the total field in the center of each of the coils be equal to a constant and then solve for the current density required in each of the coils to make this true. The same procedure could be used with a single long solenoid where the solenoid could be broken up into an arbitrary number of pieces.

The most useful method of solution of this type of problem is basically the following: taking any single coil we first find the field at the center of that coil and at distances away from the coil represented by the center lines of each of the other coils, for a given current. Then by properly adding up the coils in matrix form and demanding that the field matrix times the current matrix be a constant we can invert the current matrix and solve for the individual currents required in each of the coils to achieve the desired results. The method of solution and a useful notation are developed in the following example. Mr. R. Bradshaw of Conesco, Inc., Arlington, brought this method of solution to our attention. Consider the following system of coils arranged so that they have a common axis. They may in general be of various sizes but for illustration consider them to be all identical with the same separation. For ease in calculation choose an even number of coils so that we have a plane of symmetry passing between coils.



Our object is to find what ratio of currents is needed in the various coils so that a uniform magnetic field is obtained throughout the interior. If the coils are thin and close together, to a graph approximation uniformity is obtained by requiring that the field at the center of each coil be the same. The field falls off rapidly on each side of the center of a sample of and we may construct a matrix the elements of which represent the contributions of the coils at their respective centers. Let Fij be the field at the center of coil i due to the field of coil j. For example, for 4 coils, 2 coils on each

side of the plane of symmetry, we have

These matrix elements may be expressed in % of an infinite solenoidal sheet of the same mean radius, or whatever one desires, since we need only ratios.

Fij Ij = Fi where Fi is the field at the center of the ith coil or

If one now specifies the  $\mathcal{F}_i$ 's the  $F_{ij}$  matrix can be inverted and the  $I_j$  matrix found.

If the current in the coils are arranged so that  $I_1 = I_4$  and  $I_2 = I_3$  then it is convenient to partition the matrix as follows:

$$f_{11} = \begin{vmatrix} f_{11} & F_{12} \\ F_{21} & F_{22} \end{vmatrix}, f_{12} = \begin{vmatrix} F_{13} & F_{14} \\ F_{23} & F_{24} \end{vmatrix}, f_{21} = \begin{vmatrix} F_{31} & F_{32} \\ F_{41} & F_{42} \end{vmatrix}, f_{22} = \begin{vmatrix} F_{33} & F_{34} \\ F_{43} & F_{44} \end{vmatrix}$$

$$i_{1} = \begin{vmatrix} I_{1} \\ I_{2} \end{vmatrix}, i_{2} = \begin{vmatrix} I_{3} \\ I_{4} \end{vmatrix}; \mathcal{F}_{1} = \begin{vmatrix} \mathcal{F}_{1} \\ \mathcal{F}_{2} \end{vmatrix} \mathcal{F}_{2}^{1} = \begin{vmatrix} \mathcal{F}_{3} \\ \mathcal{F}_{2} \end{vmatrix}$$

$$\begin{vmatrix} f_{11} & f_{12} \\ f_{21} & f_{22} \end{vmatrix} \begin{vmatrix} i_{1} \\ i_{2} \end{vmatrix} = \begin{vmatrix} \mathcal{F}_{1} \\ \mathcal{F}_{2} \end{vmatrix} \text{ and then } i_{2} = ki_{1} \text{ and } \mathcal{F}_{2}^{1} = k\mathcal{F}_{1}^{1}$$

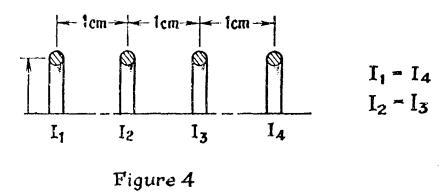
$$\begin{vmatrix} f_{11} & f_{12} \\ f_{21} & f_{22} \end{vmatrix} \begin{vmatrix} i_{1} \\ i_{2} \end{vmatrix} = i_{2} \text{ and } \begin{vmatrix} f_{11} & f_{12} \\ f_{21} & f_{22} \end{vmatrix} \begin{vmatrix} i_{1} \\ ki_{1} \end{vmatrix} = \begin{vmatrix} \mathcal{F}_{1} \\ k\mathcal{F}_{1} \end{vmatrix}$$

it follows that 
$$(f_n + f_{12}k)i_1 = \mathcal{F}_1^1$$

Since we desire the fields at the centers of each coil to be a constant 
$$\mathcal{F}_1 = \mathcal{F}_2 = \text{constant} = C$$
 where  $C = C \begin{bmatrix} 1 \\ 1 \end{bmatrix}$  i.e.,  $(f_{11} + f_{12} k)i_1 = C$ 

It is clear that  $i_1 = (f_{11} + f_{12}k)^{-1}C$  and this gives us the desired ratio of currents.

As an example consider the field on the axis due to a system of four current loops of radius y = 1cm



Separation between centers of the coils is 1 cm. We need calculate the field of one loop at the center and 1, 2, and 3 cm. on the axis away from the center. For a single loop

$$H_{X} = \frac{2\pi i}{10} \frac{y^{2}}{(x^{2}+y^{2})^{3/2}}$$

$$X | H_{X}/i = 0 .628$$

$$1 .222$$

$$2 .0562$$

$$3 .0199$$

then 
$$f_{11} = \begin{vmatrix} 0.628 & 0.222 \\ 0.222 & 0.628 \end{vmatrix}$$
  $f_{12} = \begin{vmatrix} 0.056 & 0.019 \\ 0.222 & 0.056 \end{vmatrix}$   
 $f_{12}k = \begin{vmatrix} 0.056 & 0.019 \\ 0.222 & 0.056 \end{vmatrix}$   $\begin{vmatrix} 0.1 \\ 1.0 \end{vmatrix} = \begin{vmatrix} 0.019 & 0.056 \\ 0.056 & 0.222 \end{vmatrix}$ 

$$f_{11} + f_{12}k = \begin{vmatrix} 0.628 & 0.222 \\ 0.222 & 0.628 \end{vmatrix} + \begin{vmatrix} 0.019 & 0.056 \\ 0.056 & 0.222 \end{vmatrix} = \begin{vmatrix} 0.647 & 0.278 \\ 0.278 & 0.850 \end{vmatrix}$$

We now need the inverse of  $(f_{11} + f_{12}k)$ 

By definition, the inverse of a matrix A with matrix elements  $a_{ij}$  is obtained if for each element  $(a_{ij})^{-1} = \frac{(-1)^{i+j}A_{ji}}{\det A}$ ; Aji is the minor of the determinant A.

If 
$$A = \begin{bmatrix} a_{11} & a_{12} \\ a_{21} & a_{22} \end{bmatrix}$$
 then  $A^{-1} = \begin{bmatrix} (a_{11})^{-1} (a_{12})^{-1} \\ (a_{21})^{-1} (a_{22})^{-1} \end{bmatrix}$ 

For 2 x 2 matrices this is very simple:

$$A^{-1} = \frac{1}{\det A} \begin{pmatrix} a_{22} - a_{12} \\ -a_{21} & a_{11} \end{pmatrix}$$

$$\det \left( f_{11} + f_{12} k \right) = (.850)(.647) - (278)^{2} = .473$$

$$\left( f_{11} + f_{12} k \right)^{-1} = \frac{1}{.473} \begin{vmatrix} .850 & -.278 \\ -.278 & .647 \end{vmatrix} = \begin{vmatrix} 1.797 & -.587 \\ -.587 & 1.367 \end{vmatrix}$$

$$\therefore \begin{vmatrix} I_{1} \\ I_{2} \end{vmatrix} = \begin{vmatrix} 1.797 & -587 \\ -.587 & 1.367 \end{vmatrix} \begin{vmatrix} 1 \\ 1 \end{vmatrix} c = c \begin{vmatrix} 1.210 \\ .780 \end{vmatrix}$$

$$I_1 = c(1.210)$$
  $I_2 = c(0.780)$ 

The current in coil 1 must then be 1.55 times greater than the current in coil 2.

Ingarden and Michalczyk<sup>(2)</sup> have also considered the problem of homogeneities over a certain length of the axis rather than strictly at the origin. They have used the requirement that the mean square deviation over a certain interval be minimized. Their notation is useful, but solution of the equations, which are transcendental in character, presents a number of problems.

G.

It is often of interest to know the field along the axis of a solenoid both inside the solenoid and outside the solenoid. The following two equations can be used to find the field along the axis of a uniform current solenoid and a radially distributed current solenoid (Bitter discs) respectively.

(i) Uniform current 
$$k = \frac{z}{a_1}$$

$$H = \frac{\pi i \lambda a_1}{5} \left[ \ln \frac{(k-\beta) + (\alpha^2 + (k-\beta)^2)^{1/2}}{(k+\beta) + (\alpha^2 + (k+\beta)^2)^{1/2}} - \ln \frac{(k-\beta) + (1 + (k-\beta)^2)^{1/2}}{(k+\beta) + (1 + (k+\beta)^2)^{1/2}} \right]$$
(ii) Radial current  $i = \frac{i_1 a_1}{p}$ ;  $k = \frac{z}{a_1}$ 

$$H = \frac{\pi i_1 \lambda a_1}{5} \left[ (k+\beta) \ln \left( \frac{\alpha + (\alpha^2 + (k+\beta)^2)^{1/2}}{1 + (1 + (k+\beta)^2)^{1/2}} - \frac{\alpha + (\alpha^2 + (k-\beta)^2)^{1/2}}{1 + (1 + (k-\beta)^2)^{1/2}} \right]$$

$$- (k-\beta) \ln \frac{\alpha + (\alpha^2 + (k-\beta)^2)^{1/2}}{1 + (1 + (k-\beta)^2)^{1/2}} \right]$$
(8)

Tabulated values of equation (7) for a wide variety of coil shapes have been published by D. E. Mapother and James Snyder in a University of Illinois circular #66 entitled, "The Axial Variation of the Magnetic Field in Solenoids of Finite Thickness". There is no available tabulation of equation (8)

Equations (7) and (8) can be written in a much simpler form utilizing the following angle notation:

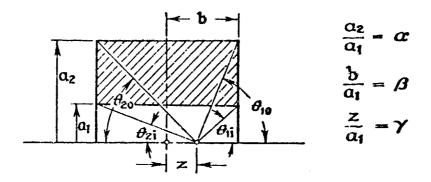


Figure 5

i) uniform current density

$$H = \frac{\pi i_1 \lambda a_1}{5} \left[ (\beta - \gamma) \ln \frac{\tan \left( \frac{\pi}{4} + \frac{\theta_{10}}{2} \right)}{\tan \left( \frac{\pi}{4} + \frac{\theta_{1i}}{2} \right)} + (\beta + \gamma) \ln \frac{\tan \left( \frac{\pi}{4} + \frac{\theta_{20}}{2} \right)}{\tan \left( \frac{\pi}{4} + \frac{\theta_{2i}}{2} \right)} \right]$$
(9)

ii) radial current density

$$H = \frac{\pi i_1 \lambda \alpha_1}{5} \ln \left[ \frac{\tan \frac{\theta_{10}}{2} \tan \frac{\theta_{20}}{2}}{\tan \frac{\theta_{1i}}{2} \tan \frac{\theta_{10}}{2}} \right]$$
(10)

General expressions for the off axis fields arising from a loop of current can be written in terms of elliptic integrals of the first and second kind. (3) They can be written in the following way, where the parameters are defined in Figure 6.

$$H_z = \frac{2I}{Q^{1/2}} \left[ F(k) + \frac{(\alpha^2 - \rho^2 - z^2)}{Q} s(k) \right]$$
 (11)

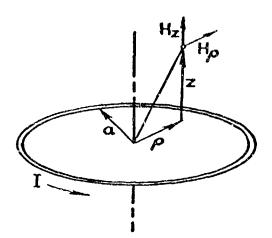
$$H_{\rho} = \frac{2I}{Q^{1/2}} \left[ -F(k) + \frac{(\alpha^2 + \rho^2 + z^2)}{Q} g(k) \right]$$
(12)

3. W. R. Smythe, STATIC AND DYNAMIC ELECTRICITY, p. 266, McGraw-Hill, New York, 1950

$$Q = (a+p)^{2} + z^{2}$$

$$k = \left(\frac{4ap}{(a+p)^{2} + z^{2}}\right)^{1/2}$$

$$S(k) = \frac{E(k)}{1-k^2}$$
 where  $E(k) =$  elliptic integral of the second kind



Using equations (11) and (12) the field at any point produced by a finite coil can be found by dividing up the coil into loops and summing all the contributions. A number of laboratories, including our own have written simple computer programs utilizing equations (11) and (12) to find the fields in and outside the conductor volumn.

On-axis and off-axis fields from any solenoid with any radial current distribution can be calculated by means of a lengthy set of tables published by the Oak Ridge National Laboratory Report #ORNL 2828, entitled, "Tables for a Semi-Infinite Circular Current Sheet," by N. B. Alexander and A. C. Downing. These tables are for the field around the end of a semi-infinite current sheet and the coil in question is built up out of these current sheets in the method shown in the introduction of the table.

Attention is also called to a tabulation by E. E. Callaghan and S. H. Maslen, NASA D465 "The Magnetic Field of a Finite Solenoid". This paper plots graphs of the radial and axial components in and around current sheets of various lengths. While not as accurate as the Oak Ridge tables, it is very useful for approximate results.

A final paper of particular relevance is that of G. R. North, "Some Parameters of Lumped Solenoids", (Oak Ridge ORNL-2975). He derives an expression for the field ripple resulting from the separation of coils in a long multicoil structure (end effects neglected). The equation is as follows (see Figure 7)

$$H(z) = H_{INE} \left(\frac{2b}{S}\right) \left[1 + \left(\frac{S}{a_1}\right)^{1/2} \frac{1}{\alpha - 1} \left(e^{-\frac{2\pi\alpha_1}{S}} - \alpha^{1/2} e^{-\frac{2\pi\alpha a_1}{S}}\right) \cdot \frac{\sin\frac{2b\pi}{S}}{\frac{2b\pi}{S}} \cdot \cos\frac{2\pi z}{S}\right]$$
(13)

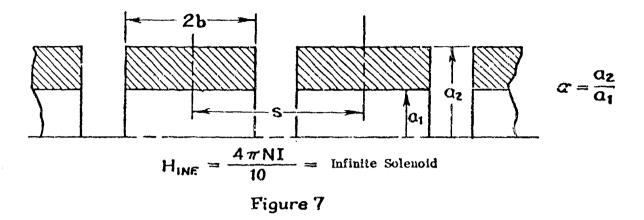


Figure 8 plots the peak value of the ripple component of eq. (13) for several values of  $\frac{2\pi z}{s}$ . The peak value occurs when z = 0,  $\cos \frac{2\pi z}{s} = 1$ .

$$\delta = \left(\frac{s}{a_1}\right)^{1/2} \frac{1}{\alpha - 1} \left(e^{-\frac{2\pi a_1}{s}} - \alpha^{1/2} e^{-\frac{2\pi \alpha a_1}{s}}\right) \tag{14}$$

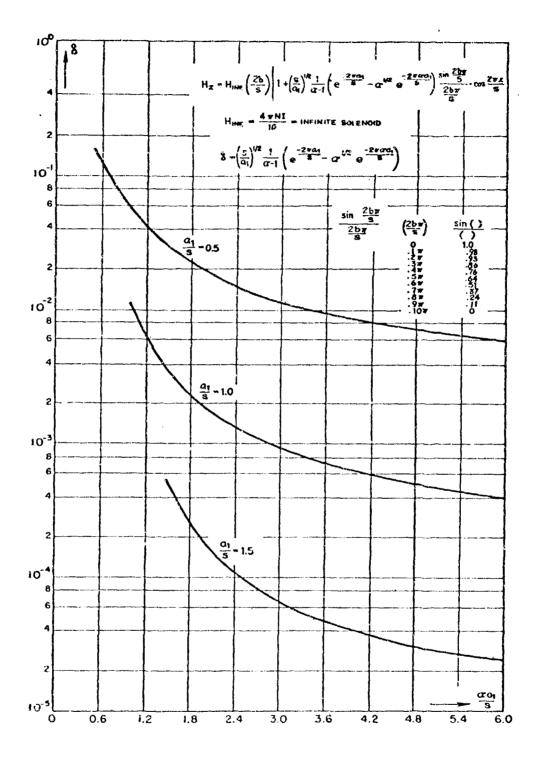


FIG. 8 PEAK VALUE OF FIELD RIPPLE IN MULTICOIL SOLENOIDS

(MULTIPLY BY 10<sup>2</sup> FOR RIPPLE IN PERCENT OF CENTRAL FIELD)

## TABLE 1 TABLE OF ERROR COEFFICIENTS

## THE & COEFFICIENTS ARE CONSTRUCTED FROM THE FOLLOWING VARIABLES:

$$c_1 = \frac{1}{1 + \beta^2} = \sin^2 \theta_{ii}$$

$$c_2 = \frac{\beta^2}{1 + \beta^2} = \cos^2 \theta_{ii}$$

$$c_3 = \frac{\alpha^2}{\alpha^2 + \beta^2} = \sin^2 \theta_{10}$$

$$c_4 = \frac{\beta^2}{\sigma^2 + \beta^2} = \cos^2 \theta_{10}$$

$$a_2$$
  $a_1$   $a_2$   $a_3$   $a_4$   $a_5$   $a_6$   $a_6$   $a_7$   $a_8$   $a_8$ 

$$\frac{a_2}{a_1} = \alpha$$

$$\frac{\mathbf{b}}{\mathbf{a_1}} = \boldsymbol{\beta}$$

$$c_5 = \ln\left(\frac{\alpha + (\beta^2 + \alpha^2)^{1/2}}{1 + (\beta^2 + 1)^{1/2}}\right) = \ln\left(\frac{\tan\left(\frac{\pi}{4} + \frac{\Theta_{10}}{2}\right)}{\tan\left(\frac{\pi}{4} + \frac{\Theta_{10}}{2}\right)}\right)$$

$$c_6 = \ln\left(\alpha \frac{\beta + (\beta^2 + 1)^{1/2}}{\beta + (\beta^2 + \alpha^2)^{1/2}}\right) = \ln\left(\frac{\tan\left(\frac{\Theta_{10}}{2}\right)}{\tan\left(\frac{\Theta_{11}}{2}\right)}\right)$$

THE COEFFICIENTS ARE USED IN THE FOLLOWING SENSE:

$$H_z = H_o \left( 1 + \mathcal{E}_2 \left( \frac{r}{\alpha_1} \right)^2 \rho_2 \left( u \right) + \mathcal{E}_4 \left( \frac{r}{\alpha_1} \right)^4 \rho_4 \left( u \right) + \cdots \right)$$

$$u = \cos \theta$$

#### TABLE 1 (continued)

CASE 1: CURRENT SHEET (I = AMP TURNS/cm.)

$$H(r_1\theta) = \frac{4\pi I}{10}\cos\theta \left[1 + \mathcal{E}_2\left(\frac{\mathbf{r}}{a}\right)^2\rho_2(\mathbf{u}) + \mathcal{E}_4\left(\frac{\mathbf{r}}{a}\right)^4\rho_4(\mathbf{u}) + \cdots\right]$$

$$\mathcal{E}_{2} = -\frac{3}{2} c_{1}^{2}$$

$$\mathcal{E}_{4} = -\frac{5}{8} \left( 4 c_{1}^{2} - 7 c_{1}^{4} \right)$$

$$\mathcal{E}_{6} = -\frac{1}{24} \left( \frac{693}{2} c_{2}^{2} - 315 c_{2} + \frac{105}{2} \right)$$

#### CASE 2: UNIFORM CURRENT DENSITY SOLENOID

$$H(o_{1}o) = G\left(\frac{W\lambda}{\rho a_{1}}\right)^{1/2} = H_{0}$$

$$H(r_{1}\theta) = H_{0}\left(1 + \mathcal{E}_{2}\left(\frac{r}{a_{1}}\right)^{2}\rho_{2}\left(u\right) + \mathcal{E}_{4}\left(\frac{r}{a_{1}}\right)^{4}\rho_{4}\left(u\right) + \cdots\right)$$

$$\mathcal{E}_{2} = \frac{1}{2\beta^{2}c_{5}}\left(c_{1}^{3/2} - c_{3}^{3/2}\right)$$

$$\mathcal{E}_{4} = \frac{1}{12\beta^{4}c_{5}}\left[c_{1}^{3/2}\left(1 + \frac{3}{2}c_{2} + \frac{15}{2}c_{2}^{2}\right) - c_{3}^{3/2}\left(1 + \frac{3}{2}c_{4} + \frac{15}{2}c_{4}^{2}\right)\right]$$

#### TABLE I

### CASE 2 (continued)

$$\mathcal{E}_{6} = \frac{1}{30\beta^{6}c_{5}} \left[ c_{1}^{3/2} \left( 1 + \frac{3}{2}c_{2} + \frac{15}{8}c_{2}^{2} - \frac{35}{4}c_{2}^{3} + \frac{315}{8}c_{2}^{4} \right) - c_{3}^{3/2} \left( 1 + \frac{3}{2}c_{4} + \frac{15}{8}c_{4}^{2} - \frac{35}{4}c_{4}^{3} + \frac{315}{8}c_{4}^{4} \right) \right]$$

$$\mathcal{E}_{8} = \frac{1}{56\beta^{8}c_{5}} \left[ c_{1}^{3/2} \left( 1 + \frac{3}{2}c_{2} + \frac{15}{8}c_{2}^{2} + \frac{35}{16}c_{2}^{3} + \frac{315}{16}c_{2}^{4} \right) - \frac{2079}{16}c_{2}^{5} + \frac{3003}{16}c_{2}^{6} \right) - c_{3}^{3/2} \left( 1 + \frac{3}{2}c_{4} + \frac{15}{8}c_{4}^{2} + \frac{35}{16}c_{4}^{3} + \frac{315}{16}c_{4}^{4} - \frac{2079}{16}c_{4}^{5} + \frac{3003}{16}c_{4}^{6} \right)$$

#### TABLE 1 (continued)

CASE 3: RADIAL CURRENT DISTRIBUTION (1=10/r)

$$H(r,\theta) = G\left(\frac{W\lambda}{\rho\alpha_1}\right)^{1/2} = H_0$$

$$H(r,\theta) = H_0\left(1 + \mathcal{E}_2\left(\frac{r}{\alpha_1}\right)^2 \rho_2(u) + \mathcal{E}_4\left(\frac{r}{\alpha_1}\right)^4 \rho_4(u) + \cdots\right)$$

$$\mathcal{E}_2 = \frac{1}{2\beta^2 c_5} \left(C_4^{3/2} - C_2^{3/2}\right)$$

$$\mathcal{E}_4 = \frac{1}{4\beta^4 c_6} \left[\frac{35}{14} \left(C_4^{1/2} - C_2^{7/2}\right) - \frac{21}{14} \left(C_4^{5/2} - C_2^{5/2}\right)\right]$$

$$\mathcal{E}_6 = \frac{1}{6\beta^6 c_6} \left[\frac{126}{16} \left(C_4^{11/2} - C_2^{11/2}\right) - \frac{140}{16} \left(C_4^{9/2} - C_2^{9/2}\right)\right]$$

$$+ \frac{3c}{16} \left(C_4^{7/2} - C_2^{7/2}\right)\right]$$

$$\mathcal{E}_8 = \frac{1}{8\beta^8 c_6} \left[\frac{6235}{240} \left(C_4^{15/2} - C_2^{15/2}\right) - \frac{7161}{24C} \left(C_4^{13/2} - C_2^{15/2}\right) + \frac{315}{240} \left(C_4^{11/2} - C_2^{11/2}\right) - \frac{525}{240} \left(C_4^{9/2} - C_2^{9/2}\right)\right]$$

$$- \frac{70}{240} \left(C_4^{7/2} - C_2^{7/2}\right)\right]$$

# TABLE OF LEGENDRE POLYNOMIALS AND THEIR DERIVATIVES (EVEN NO'S)

$$\rho_0$$
 (u) = 1

$$u = \cos \theta$$

$$\rho_0'(u)=0$$

$$\rho_2 (u) = \frac{1}{2} (3u^2 - 1)$$

$$\rho_2'(u) = \frac{1}{2}(6u)$$

$$\rho_4 (u) = \frac{1}{8} (35u^4 - 30u^2 + 3)$$

$$\rho_4'$$
 (u)= $\frac{1}{8}$  (140 u<sup>3</sup> - 60 u)

$$\rho_6 (u) = \frac{1}{16} (231u^6 - 315u^4 + 105u^2 - 5)$$

$$\rho_6'(u) = \frac{1}{16} (1386 u^5 - 1260 u^3 + 210 u)$$

$$\rho_8 (u) = \frac{1}{128} (6,435 u^8 - 12,012 u^6 + 6,930 u^4 - 1,260 u^2 + 35)$$

$$\rho_8'$$
 (u)= $\frac{1}{128}$  (51,480 $u^7$ -72,072 $u^5$ +27,720 $u^3$ -2,520)

SPACEING OF HELMHOLTZ PAIRS

TABLE 3

TABULATED VALUES OF K VS X AND A

USED TO DETERMINE PROPER SPACING BETWEEN HELMHOLTZ COILS FOR NO 2ND DERIVATIVE AT THE ORIGIN. (UNIFORM CURRENT)

A. 2(01-01)

	8.	94533	94529	94450	04230		200	94000							1					Pod in				
		9(275	90246	90161	1,000		17900	89586	83300	06000	2000	88800			T					Cor Meen Rodina	.1			
	1:0	86118	66085	85987	\$5825	100	70000	85321	84990	84810		94200					•	ſ	-		1	I		
	1 3	20179	82(164	81951	817.63	*11° 06	20040	81181	80796	802.59	100	25	79350	78600			<u> </u>	L		5	-	_[	_]	
-	1.0	0.707	78202	78071	77855	77558		77184	76739	76233	25.00	*1001	15074	74445	2000	2000		Γ		_ <u>_</u>	  - 	_[		Ţ
-	7.5	000	74513	74369	74121	73780		13.130	72938	72254	41660	0007/	70915	70183	27.74		98700			_	-	_[		Ļ
1 2	2.002	100	(1035	70864	70582	70192		20, 80	69113	68442	87508	000	56895	36050	1		200	63500	T	T		T	T	7
-	87875	04440	2/3/	67576	67258	65814	2000	00700	65587	64820	83967		63044	62068	19019		2000	54050				T	1	
0.1	64814	- 17273	16120	64524	64164	63668	2000	2000	62280	61410	60433	2	23382	58260	87098		77860	54760	†		T	T	1	
9.9	62047	61066	30.1	61724	61324	60768	6008	3	59215	58233	57136	2000	00000	54635	63319		200	50613	49320				<del> </del>	
8.0	59547	59458		59192	58750	58135	57252		56411	55318	54086	50502	35133	51280	49735	1		46631	45120	43710				1
0.7	57325	57228	3000	25938	56455	55783	54925		53883	52680	51313	40809	2005	48169	46442	44658		42656	41096	39440	38000	T		1
9.0	55330	55285	000	2,63,6	54450	53723	52793		94010	E0344	48843	47174		₹535R	13420	47.897		38336	37296	35351	33590	32100	<u> </u>	1
6.9	53746	53535	00000	23300	52744	51965	50968	1	#0 / D#	48329	46701	44890		12834	PC104	38470		33	33786	315.0	29410	27600		
4.0	.52399	52281	5,000	272.5	51341	50518	45461	40.170	7/101	46653	44909	42849		40786	38439	35937	1	33266	30652	28008	25505	23300	21600	
0.3	51350	51227	50860	2000	50246	49337	48281	480.90	10353	45332	43491	41412		39014	36580	33854	00000	20202	28003	24981	22028	18300	17000	
0.2	.50600	50474	50005		48463	48578	47434	46035	2001	44378	42463	40293		37872	35210	32321	20000	27727	25969	22597	19195	15590	12900	
0.1	.50160	50022	49638	2000	コカカカ	48085	46924	45495		43801	4184C	35613	00120	02.1.20	34368	31386	06136		24681	21054	17297	13480	06960	
0.0	.50000	49571	.49483	40004	CC221.	.47926	.46753	45315	0000	8005	41632	.35385	2000	50000	.34085	31043	27755	3	24239	.20518	.16622	.12585	.08444	
×/	00	0 1	0.5	-	2	-	\$ .	9 0		;	က ဝ	ъ О	-			 			7-1	1.5	1.8	1.7	1.8	

AXIAL ERROR LIMITS TABLE 4 TABULATION OF V VS X AND A

 $A = \frac{2(a_2 - a_1)}{a_2 + a_1} = \frac{2(a - 1)}{a + 1}$ 

9 7

> WHERE V=Z/a, AND WHERE Z REPRESENTS THE POINT AT WHICH THE FIELD DIFFERS FROM THE CENTRAL FIELD BY O.1 % (UNIFORM CURRENT)

1.7 1.8	290 .302	. 582	288	286						-		_		14	a» * Mean	Radius	
1.6	.278	277	276	274	172	268						. ء		$\int$			-
1.5	.266	266	265	263	260	257	252	247				- 6	1			^	1
1.4	.255	254	253	252	249	246	242	237	231	225						(	
1.3	.244	244	243	241	238	235	231	227	221	215	208			+	,		
1.2	.234	234	233	231	228	225	222	217	212	206	159	191				<b>-</b>	1
1.1	.224	224	223	221	219	216	212	208	203	197	190	182	174				<u></u>
1.0	.216	215	214	212	210	207	203	199	194	188	181	174	166				
0.9	.207	207	206	204	202	159	195	191	186	180	173	166	158	149			
9.0	.200	200	199	197	195	192	188	183	178	172	166	159	151	142	132		
0.7	.193	193	192	190	188	185	181	177	111	165	159	152	144	135	125		
0.6	.188	187	186	185	182	179	175	170	165	159	152	145	137	128	119		
0.5	.183	182	181	180	177	174	170	165	160	154	147	139	131	122	112		
0.4	.179	178	177	176	173	173	997	161	155	149	142	134	126	116	106	960	
0.3	941.	175	174	172	170	166	. 162	157	152	145	138	130	121	111	101	060	
0.2	.173	173	172	170	167	164	160	155	149	143	135	127	118	108	097	085	
0.1	.172	172	171	169	166	163	158	153	148	141	133	125	116	105	094	082	
0.0	.172	.171	170	.168	.166	.162	.158	.153	.147	.140	.133	.124	.115	.104	.093	.081	.067
X/	0.0	6.1	0.2	0.3	0.4	0.5	ં.6	0.7	0.8	0.9	1.0	1.1	1.2	1.3	1.4	1.5	1.6

AXIAL DISPLACEMENT ERROR COEFFICIENT Q FOR IMPROPEPLY SPACED HELMHOLTZ COILS (UNIFORM CURRENT) H = H<sub>0</sub>  $\left(1+Q\left(\frac{d}{d_0}\right)\left(\frac{z}{d_0}\right)^2+\mathcal{E}_4\left(\frac{z}{d_0}\right)^4+\cdots\right)$ TABLE 5

× 4
-----

1.7	1.95	1.96	1.59	2 03	2.09									1	°	4
1.6	2.02	2.03	2.05	2.10	2.16	2.26	2.37						•			
1.5	2.09	2.10	2.13	2.17	2.24	2.33	2.46	2.62				민		<u>.</u>		1
1.4	2.17	2.18	2.21	2.26	2.33	2.42	2.55	2.71	2.93	3.21	2.61	Î	1	~		1
1.3	2.27	2.28	2.31	2.36	2.43	2.52	2.65	2.82	3.04	3.33	3.71					
1.2	2.38	2.39	2.42	2.47	2.54	2.64	2.77	2.94	3.17	3.46	3.85	4.38	5.10	L		$\dashv$
+:1	2,51	2.52	2.55	2.60	2.67	2.77	2.91	3.08	3.31	3.62	4.02	4.57	5.33			
1.0	2.64	2.65	.2.68	2.74	2.81	2.92	3.06	3.24	3.48	3.79	4.21	4.78	5.57			
6.0	2.79	2.80	2.83	2.89	2.97	3.07	3.22	3.41	3.68	3.99	4.42	5.02	5.84	7.05		
 8.+	2.85	2.96	2.99	3.05	3.13	3.24	3.40	3,59	3.86	4.20	4.66	5.29	6.16	7.43	9.36	
++	3.11	3.12	3.15	3.21	3.30	3.42	3.58	3.79	4.07	4.43	4.92	5.58	6.51	18.7	86.6	
9.0	3.27	3.28	3.31	3.38	3.47	3.60	3.77	3.59	4.28	4.67	5.19	5.91	6.91	8.37	10.6	
+	3.42	3.43	3.47	3.54	3.63	3.77	3.95	4.18	4.50	4.91	5.47	6.24	7.34	8.93	11.4	
°, +	3.56	3.57	3.61	3.68	3.73	3.92	4.11	4.36	4.70	5.14	5.75	6.58	7.77	9.54	12.3	16.8
 6	3.68	3.69	3.73	3.80	3.91	4.06	4.26	4.52	4.87	5.35	5.99	6.89	8.19	10.1	13.2	18.4
-2.5	3.77	3.78	3.82	3.90	4.01	4.16	4.37	4.64	5.01	5.51	6.19	7.15	8.55	10.7	14.1	20,1
+ 0.1	3.83	3.84	3.88	3.96	4.07	4.22	4.44	4.72	5.10	5.61	6.32	7.32	8.79	11.1	14.8	21.4
0.0+	3.84	3.86	3.90	3.98	4.09	4.24	4.46	4.75	5.13	5.64	6.37	7.38	8.87	11.2	15.0	21.9
×/	0.0	0.1	0.2	0.3	4.0	0.5	9.0	0.7	8.0	6.0	1.0	1.1	1.2	1.3	1,4	1.5

g = Proper Spacing d = Displacement Error

ao \* Mean Radius

FOURTH ORDER ERROR COEFFICIENTS  $\mathcal{E}_4$  FOR PROPERLY SPACED HELMHOLTZ COILS. (UNIFORM CURRENT) H = H<sub>o</sub>  $\left(1 + \mathcal{E}_4\left(\frac{z}{G_0}\right)^4 + ...\right)$ 

X a a + a .

Γ		7	_	^,	T.	$\prod$	_1	_		T	T		Τ	7	7		7	$\top$		<u> </u>	8	_
	1.7	1	.142	.142		C#1.	.149	.153											ſ	-	<u>a</u>	ļ
	1.6	1	.168	.169	2.0	7117	17.	.184	.194	200	107:									+		
	1.5	-	.200	.201	200	5 3	277	219	.231	246	02.3	.267				*		1	L			,
	1.4	1	.237	.239	243	2 2	267.	.260	.274	995	004	.318	350		385	.452		7.	*r			
	1.3	1	787.	.284	288	200	167:	.309	.326	349		.379	417		.408·	.536		٦٩		+	Q,	
	1.2	100	£00°	.336	342	959	3000	.367	.388	415		451	.498	0.55	noc:	.641	753	2 6			20	
	7:	106	Coc.	.397	.404	417	10,	.435	.460	.493	6	155.	.594	055	500.	.769	706	9	20.1	T		
	1.0	464	FOE:	.466	.475	490	2	CT	.544	.584	000	.035	.708	008	3	.922	1.09	33				
	6.0	541	ָּ֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓֓	.544	.555	.573	603	100.	.538	.688	757	¥01.	.840	954		1:11	1.31	1.60	30			
	0.8 1	625		.629	.642	.664	809	000	. (43	.805	n a a	3	.991	1.13		1.32	1.58	1.94	2 47	3 97	;	
	0.7	.714		617.	.735	.762	809		,00.	.931	1 03	2	1.16	1.33		70.7	1.89	2.35	3.02	4.05		_
	3.1	.805	5	110.	.830	.862	909	070	016.	1.06	1.18		1.34	1.56	200	60.1	2.26	2.84	3.71	5.05		1
	رن ا	.895	000	206.	.923	.961	1.02	5	20:1	1.20	1.34		1.53	1.79	216	21.17	2.66	3.41	4.53	6.31		1
	ا خ	.978	900	0000	1.01	1.05	1.12	1 91		1.33	1.49		1.72	2.03	9 46	2	3.09	4.03	5.48	7.85	12.0	
0		1.05	1 08		1.09	1.13	1.20	1.30		1.44	1.63		7.89	2.25	97.6	2	3.51	4.66	6.49	9.62	15.3	
0.0	7, 1	1.10	17.		1.14	1.20	1.27	1.38		1.53	1.74	6	2.02	2.42	3.00		3.87	5.22	7.45	11.4	19.1	
6	5 1	1.14	1.15		1.18	1.23	1.32	1.43		1.59	1.81	.:	71.7	2.54	3.17		4.11	5.62	8.16	12.8	22.4	1
0	3 1	1.15	1.16		1.19	1.25	1.33	1.45	-	10.1	1.83	5.	17:7	2.58	3.23	18	4.20	5.76	8.42	13.4	23.7	
×		0.0	0.1		0.2	0.3	9.0	0.5	9	0	0.7	α		6.0	1.0	ł	7 7	1.2	1.3	1.4	1.5	

TABLE 7 SIXTH ORDER ERROR COEFFICIENTS  $\mathcal{E}_6$  FOR PROPERLY SPACED HELMHOLTZ COILS (UNIFORM CURRENT) H = H<sub>o</sub> (1+...+ $\mathcal{E}_6$ ( $\frac{z}{d_o}$ ) +...)

1.7	.135	.137	.145	.159	.181											
1.6	.160	.162	.172	.187	212.	.247	.299									
1.5	.189	.192	.203	.221	.250	.292	.352	.439								 
1.4	.225	.228	.241	.262	.296	.345	.415	.518	.670	.898	1.22			_	_	_
1.3	.268	.272	.286	.311	.351	.408	.490	.611	.789	1.06	1.48	_				_
1.2	.318	,323	.340	.369	.415	.482	.579	.720	.929	1.25	1.75	2.57	4.00			_
1:1	.379	.384	.403	.438	.490	.568	.680	.845	1.09	1.46	2.05	3.02	4.71		_	
1.0	.449	.456	.478	.517	.577	999.	.796	.987	1.27	1.71	2.40	3.54	5.55			
6.0	.530	.538	.563	.607	.675	777.	928.	1.14	1.47	1.98	2.78	4.13	6.50	11.0		
8.4	.621	.630	.658	707.	.784	888	1.06	1.31	1.68	2.26	3.18	4.75	7.54	12.9	24.1	
2.0	.722	.731	.752	.816	.901	1.03	1.21	1.48	1.89	2.54	3.58	5.36	8.60	14.9	28.4	
9.0	.828	.838	.871	.930	1.02	1.18	1.36	1.65	2.10	2.79	3.94	5.92	9.58	16.9	32.8	
6.5	.935	.946	.982	1.04	1.14	1.28	1.49	1.80	2.27	3.01	4.22	6.33	10.3	18.4	36.9	
4.0	1.04	1.05	1.09	1.15	1.25	1.40	1.62	1.93	2.51	3.15	4.38	6.53	10.6	19.2	39.4	94.2
0.3	1,13	1.14	1.18	1.25	1.35	1.50	1.72	2.04	2.51	3.24	4.42	6.49	10.4	18.8	39.0	97.0
0.2	1.20	1.21	1.25	1.32	1.43	1.58	1.80	2.11	2.57	3.26	4.37	6.26	9.79	17.1	35.0	87.6
0.1	1.25	1.26	1.30	1.37	1.47	1.63	1.84	2.15	2.59	3.26	4.29	6.01	9.08	15.2	29.4	60.1
0.0	1.26	1.27	1.31	1.38	1.49	1.64	1.86	2.16	2.60	3.25	4.25	5.90	8.77	14.4	26.8	60.0
×/	10			0.3	4.0	0.5	0.6	0.7	0.8	6.0	1.0	1.1	1.2	1.3	1.4	1.5

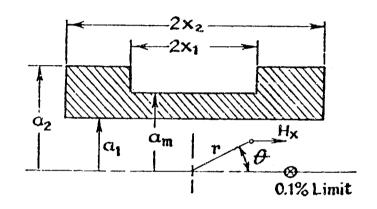
TABLE 8

Typical 6th order solenoids (No 2nd or 4th derivative on axis) where length is fixed.

(W. Garrett)

a <sub>1</sub>	x <sub>1</sub> /3 <sub>2</sub>	*2/a2	$\mathcal{E}_{6 \times 10^{-3}}$	E <sub>8</sub> x 10 <sup>-3</sup>	0.1 % Limits*
1.0	1.39327	1.82234	-5. 504	-2.634	75
0.9	1.41768	1.82451	-4.918	-2.56 <b>6</b>	77
0.8	1.44747	1.82933	-4. 255	-2.403	79
0.7	1.48310	1.83709	-3.551	-2.151	81
0.6	1.52472	1.84784	-2.849	-1.831	84
0.5	1.5719 <b>3</b>	1.86124	-2.194	-1.478	88

\* % of a along axis



$$a_2 \equiv 1.0$$

$$a_m \equiv \frac{a_2 + a_1}{2}$$

$$u = \cos \theta$$

$$H_{X} = H_{O} \left[ 1 + \mathcal{E}_{6} \left( \frac{\mathbf{r}}{a_{2}} \right)^{6} \rho_{6} \left( \mathbf{u} \right) + \mathcal{E}_{8} \left( \frac{\mathbf{r}}{a_{2}} \right)^{8} \rho_{8} \left( \mathbf{u} \right) + \cdots \right]$$

TABLE 9

Sixth Order Long Solenoids: at = 6 cm, az = 12cm, b = 36cm

Cancelled Second & Fourth Derivatives ( $H=H_0(1+\mathcal{E}_Z^z+\mathcal{E}_4z^4)$ )  $\mathcal{E}_Z=.607\times10^{-2}\%$ ,  $\mathcal{E}_A=.640\times10^{-5}\%$ 

 $i\sigma_{\gamma}^{1}: \alpha_{1}*6cm$  Current sheet  $\alpha_{2}*12cm$  on I.D. on O.D. b \*36cm

Uniformly Two separate reduced current: sheets placed at in central region ends on I.D.

One notch on O.D.  $\frac{1}{1} = +207$ 

(+)0.27%

I = -207
(-)0.45%
$ (.536 \times 10^{-9})_{2}^{6} (\%)   (.759 \times 10^{-6})_{2}^{6} (\%)   (.303 \times 10^{-8})_{2}^{6} (\%)   (484 \times 10^{-6})_{2}^{6} (\%)   (1.84 \times 10^{-4})_{2}^{6} (\%)   (.304 \times 10^{-4})_{2}^{6} (\%) $

(.758×10°) z\*(%)

****
× 69°,49°,
i i

20 a a a /a i B a b /a i i e Constant

-0.53922 CG -0.49108-CG -0.46398-CG -0.3300E-CO	A2 -0.6432E 00			A6 -0.68716-01	48 0.4245E-00	62 0.14085-60	6272 -0.9653E-01	6244 0.4035£-01	\$286 -0.9671E-02	62A8 0-5975E-01
0.20026-00 - 0.18756-00 - 0.15436-01	0-22486-00		-0.530	7E-01	0.31676-00	0.15C8E-00	-0.81316-01	0.33726-01	-0-8004E-02	0.47765-01
C.18436-01	0.20026-00		-6.47678	70-	0.28016-00	0-15186-00	-0.74545-01	0.39392-01	-0.71465-02	0.4253E-01
10-37.4.0	0.18756-00		6.202	3 6	0.26195-00	0.15075-50	-0.6991E-C1	0.2825E-01	-0.6635E-02	0.39476-01
10101010	0-18165-01		0.14235-	, -, ; ;	0.62025-01	0-18735-00	-0.54445-01	0.36026-02	0.34446-02	0.1447E-01
0.16386-01	0.16386-01		C-12366-0	-4	0.54226-01	0.19C9E-00	-0.5134E-01	0.32412-02	0.2360E-02	0-10356-01
0.16035-63	0.16035-63		0-114/5-	7	0.5035E-01	0.19C8E-C0	-0.48348-01	0.30595-02	0.21886-02	0.96056-02
-0-91356-05	-0-32516-0-		-0.3720£-C	ţ	0.19475-02	0-16935-00	-0.1362E-C1	-0.1554E-02	-0.62976-05	0.32966-03
-0.6.9.6-02	-0.6.976-02		0.12325-0	5	0.1487E-02	C.1957E-60	-0.16296-01	-0-12135-02	0.23526-04	0.2911E-03
-3-613-6-01 -0-493/6-02	70-3776-0-		0-10-01-0	٦.	0-12/56-02	0-36402-0	-0-1666E-C1	-0.1012E-02	0.2245E-04	0.26135-03
-0-33:76-02	-0-4360E-02	ŧ	0-78184.0		70-30217-0	0.20825-00	-0.16436-01	-0.90886-03		0.2+285-03
-0.2044.6-02	-0.2044.6-02		-0.79366-0		0.6774[-04	0.18301-00	-0.54276-07	-0.37406-03	40-30017-0-	10-1640110
-0-11011-0-	-0-11011-02		7-39684-0-	4	0.6185[-04	0.1956-00	-0.6159E-C2	-0.33425-03		0-1210F-14
-0.14386-02	-0.1438E-02		-0.4773E-C		0.56316-04	0.20166-00	-0.64325-02	-0.30206-03		0.11356-24
-0.6701E-03	-0.6701E-03		-0-33246-0		0.4533E-US	0-13866-00	-0-1314E-C2	-0.92895-04	-0.4650E-05	0-62336-06
**************************************	-0.6865E-03		-0-2461E-0		0.5984E-05	0.16846-00	-0.2119E-02	-0-1156E-03		0.1006E-US
	-0.56136-01		0 - 0 - 0 - 0 - 0 - 0 - 0 - 0 - 0 - 0 -		0.37365-03	0.14275-00	-0.26326-02	-0.11435-03		0-1088E-05
2207E-03	2207E-03		-0-8620E-0	rs	0.43126-05	0-12646-00	-0.5424E-02	-0.1068E-03	-0.2902E-05	0.1062E-05
-0.25776-03	-0.25776-03		-0-11166-0	_	0.74536-06	C-1555E-00	-0.94476-63	-0-4609E-04		0-11605-05
-0.2569E-03	-0.2569E-C3		-0-62535-0	۰.	0.858CE-C6	0.17036-00	-0.1257E-C2	-0-43756-04		0.1478E-06
	-0-24145-03		-0.51745-05	_	0.8717E-06	0.1788£-00	-0-1473E-02	-0-4317E-04		0-15596-06
-0-10-10-0-	-0-10-10-0-		-0.26.516-0	٠.	0-12045-07	0.14.701-55	-0.2557E-03	-0.9871E-05		0.65056-03
-0.1258E-03	-0.1258E-03		-0.2354L-0	~	0.1635t-06	0.15946-00	-0.6550E-03	-0-1997E-04	-0.335E-00	0-1/385-07
	-0-11476-03		-0.20056-0	<u>.</u>	0.18036-08	0.16841-00	-0,8023E-03	-0-19305-04	-0.3375E-04	0.3035E-07
-0-12616-02 -0-36246-04 -0-83786-06 -0-18636-02 -0-6-6936-04 -0-69366-04	-0.5629E=04		-0-6578E-0	٠,	0.8950E-08	0.10926-00		-0.3962E-05		0.97836-09
-0.56356-04	-0.56356-04		-0-9626E-0		0.36846-07	0-13006-00	-0-3255-03	CO-07/190.0-	-0.1353E-06	0.31876-38
-0.5861E-04	-0.5861E-04		-0-3000E-0	•	0.44836-07	0.15916-00		-0.93236-05		D. 7.335-08
-0.17196-04	-0.1719E-04		-0.3256E-0	٠ و	0.16945-08	0-1026E-00		-0.1764E-05		0.17386-09
0 - 1 - 1 - 1 - 1 - 1 - 1 - 1 - 1 - 1 -	40-365 37-0-		0-44-07-0-	٥.	0.5365E-08	0.12756-00		-0.3117E-05		0.68596-09
	-0-31765-04		-0-4440E-0-		1979610	00-14196-00	-0.2174E-03	-0.4150E-05		0.13506-08
10-31040-0-	10-31040-0-		10.3357510		10-37171-0	0.136361.0	-0.2834E-03	-0-47506-05		0-1919E-08
10-10-0-0-0-	10-10-0-0-0-		70-376610-		0.33346-09	C. 9709E-01	-0.4513E-04	-0-4513E-04 -0-8567E-0E		0.34366-10
-0-1605E-C4	-0-1605E-C4		0-19746-0-	<b>,</b>	0.27455-00	00137771	40-94040-0-		-0.2176E-37	0.1678E-09
-0.1835E-04	-0.1835E-04		-0.1952F-(	9	0.3997F-08	0.14485-00		-0-2 04E-05	-0.2661E-07	0.3701E-03
-0-4807E-05	-0-4807E-05		-0.6134E-		0.76055-10	C. 9236E -03	-0-7849F-04	**************************************	10-36087-0-	0-344E-08
-0.7216E-05	-0.7216E-05		-D.8458E-	07	0.3940E-09	0-11545-00	-0.5692E-04 -0.8329E-06	-0.83295-06		0.025
	-0.9238E-05		-0.3745E-	5	0.56796-09	0.1286E-00	-0.8796E-04	-0.1188E-05		0.11166-09
-6.8673E-03 -0.1070E-04 -6.1006E-06	-0-10706-04		-6.10066-	9	0.13666-08	0.13746-00	-0-1191E-03			0-18766-09

TABLE 10B
ERROR COEFFICIENTS AND GEOMETRY FACTORS FOR RACIAL CURRENT DENSITY  $j = \left(\frac{1}{2}, \frac{1}{2}, \frac{1}{2}\right)$ VS. a AND  $\beta$ H = H<sub>e</sub>  $\left(i + A_2\left(\frac{z}{a_1}\right)^2 P_2(u) + A\left(\frac{z}{a_1}\right)^4 P_4(u) + \cdots\right)$ 

TABLE 11 A SECOND ERROR COEFFICIENTS AND GEOMETRY FACTORS FOR SMALL  $\beta$ 's.  $(0.1 + \beta - 1.0)$  UNIFORM CURRENT  $+ H + H_0(1 + D_2(\frac{1}{01})^2 P_2(u) + \cdots)$  RADIAL CURRENT,  $H + H_0(1 + A_2(\frac{1}{01})^2 P_2(u) + \cdots)$ 

0.583986-01 -0.58386-01 -0.44386-01 -0.44386-01 -0.44386-01 -0.44386-01 -0.44386-01 -0.44386-01 -0.58436-01 -0.58436-01 -0.584386-01 -0.584386-01 -0.584386-01
6.16.2 -6.50220)E-01 -0.54035E-01
62 0.45642E-01 0.461446E-01 0.76167E-01 0.76167E-01 0.76167E-01 0.77168E-01 0.177168E-00 0.177168E-00 0.177168E-00 0.177168E-00 0.177168E-00 0.177168E-00 0.177168E-00 0.177168E-00 0.177168E-00 0.177168E-00 0.177168E-00
C1 C450034E-C1 C46406E-C1 C46406E-C1 C77341E
-0.12728E 01 -0.12728E 01 -0.374539E 00 -0.345439E 00 -0.35753E 00 -0.457543E 00 -0.457543E 00 -0.457543E 00 -0.457544E 00 -0.457544E 00 -0.457544E 00 -0.457544E 00
0.1710/E 01 0.1710/E 01 0.7170/E 01 0.7170
00000000000000000000000000000000000000

TABLE II B FOURTH ERROR COEFFICIENTS AND GEOMETRY FACTORS FOR SMALL B's. (0.1-8-2.0) UNIFORM CURRENT: H:  $H_0(1+\cdots+D_0(\frac{z}{4})^4P_2(u)+\cdots)$  RADIAL CURRENT: H:  $H_0(1+\cdots+D_0(\frac{z}{4})^4P_2(u)+\cdots)$ 

A: P×4	4 4574	ć	;				
		***	Ç.	3	62	4010	463
	000	10 364507110	0-128751E 01	0.4580346-01	0.4585426-01	0.543115-01	0.400
3 4 4 7	00-404-0	0-1262856 00	0-7276586 00	0.6141036-01	0.4147486-01	10-11-11-11-11-11-11-11-11-11-11-11-11-1	
0.115	1 0.43E CO	0.250530E-00	C-7-3630F-00	0.2000414-01	70 100 100 100	10-31-5-0	***
0.418.0	1 0.8CE 00	0-5494105-01	101111111111111111111111111111111111111	70.078600780	TO-3/7010/-0	10-30-0061-0	C-196
3-115	A LOCK CO	-0-2485616-61	101111111111111111111111111111111111111	70-307474747	0. F42C52E-01	0-4073416-02	0.405
0.110	1 7 126 01	40 - 34 C C C C C C C C C C C C C C C C C C	10-240/052-0-	0-1242226-01	0.7551438-01	-0-1875765-02	-0.189
	1710	10-317-604-0	10-34:5954-5-	0.7516036-01	0.7521406-01	-0-3643365-02	40.40
9 4 4 4 4 4 4 4 4 4 4 4 4 4 4 4 4 4 4 4	10 244-0	-0-3551535-0-	-0.3557725-02	0.7355755-01	0-7401546-01	-0-2424975-03	
241.3	10 a-1-0	-0-2022020	-0-2653406-01	6.7233647-01	0-174076-01	20-14/01/01/04	
0.11E G	10 331-0 1	-0-1846106-01	-0-106511E-01	0.5043150-03	0.3642146.04	70-350374745-	761-01
	10 302.0 1	-0-120040E-01	10-1242745-01	10-10-10-10-0	10-301-501-0	-0-1313131-02	-6-131
U 306 °0	0.205-00	0-1540536-06	0.175.000	10-316243940	10-3707CB00	-C-887477E-03	-0.653
_		D-7470746-00	001404444	TO-Tracegoen	10-3508866-01	0-3143335-01	0.478
0.00	0 NC 4 C	00 000000000000000000000000000000000000	00104001040	00-174477-00	0.1374425-00	0-2866785-01	0.423
	200	004377917799	00-34000000	0-141226E-00	0.1414105-00	0-1805165-01	0.753
	20 270	0.57755	G-423432E-01	0-1260911-0	0.1773665-03	0.3912676	
0 0 0 0 0 0	00 9999	10-3919802.0	C-121C2C2-01	0.1660915-00	0.1873415-00	10.3181816.0	
0.308	7 C-12E C1	0.451175E-02	-0.304625E-03	3-1725616-00	0.1444441.0	70-31710000	2
0.30E	1 C.14F C1	-0.219365E-02	-0-6601758-02	00-34747-00	DO-13707010	U-1/422/E-03	-0.56
0.306 0	1 0.155 01	-0-4-8766F-03	70 30 40 40 60 60	00134940140	C- 170308E-00	-G-3671036-03	-0.129
6,328.0	1 C.125 G1	CC-10/10/10/10	7019773184401	20-12-20-20	0-1973465-00	-0-800593E-03	-0-155
10 1 1 1 1 C		70131/35754	70-36T-367-31	0-179172-00	0.197053E-00	-0-8773476-03	-0-143
	100000000000000000000000000000000000000	70-3:000000	-C-914630E-02	0.1763612-00	G.135741E-CO	-0.815721F-03	-0-131
4.4.5	201924	0.2107376-00	C-400334E-00	C-732224E-01	0.9915096-01	D.183742E-01	0.197
		05-3775671-0	0-2546725-00	0-101821E-00	0-1370196-00	0-1645375-03	
	2000	10-30226	C.129643E-00	0.121531£-00	0-162012E-00	C-107595F-01	0.5
		10-343-404-1	0-529574E-61	C-135873E-03	0-1791176-00	0.5469036-02	
		4-155551-01	0.1603486-01	0-146445E-00	0-190760E-00	0.2144816-03	
3000		0-4212216-02	0.8730492-03	0.154235E-00	0-1985095-00	20-3194646 0	
20.50	10 341.0 1	0-1872246-03	-0-4330236-02	0.1599246-00	0-201688	0 14554	
0.508.0	1 0.165 02	-0-153932E-02	-D. 544.7618-02	0.1440.78-00	100000000000000000000000000000000000000	0-543561E-04	1000
C. 50E 0	1 0,168 02	-0.2031796-02	-0-412476-03	00-11-10-1-1	00-1215-00	-0.2524746-05	-0-112
0.308.0	1 9.205 01	-60-30 B. V. F. C. O. S.	**********	2012/00/00/00	0.20/813E-00	-0.339034E-03	-0.10
		47:41:44	30-3:0005550	00-3/6/10/10	0.206173E-00	-U-342248-U-	-0-404

-0.525566E-01 -0.525566E-01 -0.502057E-01 0.301150E-02 0.412099E-02 0.335101E-02 62 0.458542E-01 0.616748E-01 0.701627E-01 0.742652E-01 0.755146E-01 0.755140E-01 C1 C.458034E-01 O.616103E-01 O.76161E-01 O.75456E-01 C.754556E-01 =  $H_0\left(1 + \cdots + D_6\left(\frac{z}{a_1}\right)^6 P_6(u) + \cdots\right)$ =  $H_0\left(1 + \dots + A_6\left(\frac{z}{a_1}\right)^6 P_6(u) + \dots + A_6\left(\frac{z}{a_1}\right)^6 P_6(u)\right)$ A6 -0.115192E D1 -0.32502E-00 0.435002E-01 0.44703E-01 0.44703E-01 0.114744E 01 -0.327958E-00 0.429624E-01 0.454103E-01 0.160882E-01 COEFFICIENTS  $\beta$  's. (0.1  $-\beta$  -3) I CURRENT . H UNIFORM CURRENT 0.20E-00 0.40E-00 0.60E-00 0.60E-00 0.00E-00 0.12E-01 ERROR TABLE 11C RADIAL SMALL SIXTH

-0.528602E-01 -0.528202E-01 -0.20203E-01 0.305209E-02 0.315814E-02 0.370912E-02 0.270917E-02 0.270917E-02 -0.852843E-04 -0.851845E-04	100.128999E-04 100.128999E-04 100.128999E-04 100.128999E-01 100.128999E-03
61Db -0.525566E-01 -0.202057E-01 0.301150E-02 0.612099E-02 0.120920E-02 0.120920E-03 0.272292E-04 -0.947023E-04 -0.858849E-04 -0.858849E-04	-0.285403E-04 -0.1230248E-04 -0.1230248E-04 -0.112687E-04 -0.112687E-04 -0.159289E-02 -0.59289E-03
62 0.458542E-01 0.416748E-01 0.701627E-01 0.74262E-01 0.755141E-01 0.755140E-01 0.74716E-01 0.645262E-01 0.645262E-01	0.629381E-01 0.9564126-01 0.9564126-01 0.9564126-01 0.1374826-00 0.1374826-00 0.1374826-00 0.1373866-00 0.1373866-00 0.1973866-00 0.1973866-00 0.1973866-00 0.1973866-00 0.1973866-00 0.197386-00 0.2034556-00 0.203456-00 0.203456-00 0.203456-00 0.203456-00 0.203456-00 0.203456-00 0.203456-00 0.203456-00 0.203456-00
C1 0.4580346-01 0.6161036-01 0.751616-01 0.7516096-01 0.7516096-01 0.7596756-01 0.7596756-01 0.7643156-01 0.6658166-01	0.6229G76E-01 0.8639G1E-01 0.8639G1E-01 0.19413E-00 0.19413E-00 0.1725G1E-00
A6 -0.115192E 01 0.320502E-00 0.435002E-01 0.16742E-01 0.344703E-01 0.344703E-02 -0.135269E-02 -0.135269E-02 -0.135269E-03	10.259986-03 -0.3668986-03 -0.3608366-03 -0.3608366-03 -0.1002096-01 0.1723066-01 0.1723066-02 0.3892376-02 0.3892376-02 0.3892376-03 0.168596-03 0.168596-03 0.168596-03 0.168586-04 0.2832396-03 0.2832396-04 0.2832396-04 0.28338256-04
-0.116744E 01 -0.327988E-01 0.429624E-01 0.625501E-01 0.1648088E-01 0.16888E-01 0.56818E-01 -0.433876E-02 -0.433876E-02 -0.433876E-03 -0.433876E-03	-0.2693627-03 -0.20693627-03 -0.20693627-03 -0.20693627-03 -0.113637-03 -0.113637-03 -0.113637-03 -0.113637-03 -0.113637-03 -0.206267
0.406 0.406 0.606 0.606 0.606 0.106 0.106 0.106 0.206 0.206 0.206 0.206	

TABLE 11D EIGHTH ERFOR COEFFICIENT ... SMALL  $\beta$  \* (0.1  $\beta$  \* 0.0) UNIFORM CURRENT:  $H \circ H_0 \left(1 + \cdots + D_0 \left(\frac{L}{d_1}\right)^8 P_0(u) + \cdots + P_0 \left(\frac{L}{d_1}\right)^8 P_0(u$ 

0 4 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7
0.137.048-02 0.137.1348-03 0.137.1348-03 0.137.1
0.3573.46 0.3 0.383.86 0.3 0.338.37 0.4 0.3 0.4 0.3 0.3 0.3 0.3 0.3 0.3 0.3 0.3 0.3 0.3
0.135721-01 0.135721-02 0.135721-03 0.1357
0.3875(E-0) 0.258815E-02 0.788815E-01 0.288218E-01 0.288218E-01 0.288218E-02 0.288218E-02 0.288218E-02 0.288218E-02 0.288218E-02 0.288218E-03 0.2882
-C.233176-C.4 0.181016-01 0.22318-01 -C.2323176-C.4 0.4810316-01 -C.2323176-C.4 0.48103176-C.4 0.48103176-C.4 0.48103176-C.4 0.48103176-C.4 0.4810316-C.4 0.48103176-C.4 0.48103176-C.
-6.22305/E-05 0.15497E-03 0.55691E-01 0.15626/E-05 0.15497E-03 0.55691E-01 0.16626/E-05 0.15497E-03 0.55691E-01 0.166299/E-01 0.15699/E-01 0.15699/E-01 0.15699/E-01 0.15699/E-02 0.15969/E-03 0.15969/E
-0.112.02.02.00.03.43.02.00.02.02.02.02.00.00.00.00.00.00.00.
C.100100E-01   C.20000E-02   C.11411E-02   C.12000E-02   C.1200E-02   C.12000E-02
-0.18308E-02 0.1286C-00 0.18409F-00 0.18409F-00 0.18408E-00 0.1840
1888E-G2 6.226.85-01 0.16668E-C0 0.22286E-C0 0.22286E-C0 0.2238E-C0 0.2238
0.402446-03 0.135917E-01 0.172481E-00 0.402446-03 0.135917E-01 0.13447E-01 0.1
0.1446.85.796-03 0.0334.86.04 0.1748.11-00 0.1748.14-00 0
C.1942681-09 C.1972117E-02 C.179111E-02 C.179111E-02 C.179111E-02 C.179111E-03 C.179111E-03 C.179111E-03 C.179111E-03 C.179111E-03 C.179111E-04 C.179111E-04 C.179111E-05 C.17
0.7712911-0* 0.184773;-0? 0.174801[-0.0.0.0.0.0.0.0.0.0.0.0.0.0.0.0.0.0.0.
C.   C.   C.   C.   C.   C.   C.   C.
0.424.84800 0.1212.00 0.11213.00 0.172138.84.00 0.17534.8400 0.17534.84.00 0.17534.84.00 0.17534.84.00 0.17534.84.00 0.17534.84.00 0.17534.84.00 0.17534.84.00 0.17534.84.00 0.17534.84.00 0.18538
0.192491-0.0 0.194891-0.0 0.1100394-0.0 0.192491-0.0 0.19
0.755476-0.0.4104016-0.0.14046-0.0.1
0.176466 06 0.246546-04 0.6652115-05 0.0996 91
-0.9965910-07 C.131399C-05 0.152835-00 0.1213919-07 C.131399C-05 0.152835-00 0.151399C-05 0.152835-00 0.151399C-05 0.152835-00
-0.2245011951-06 6444228-05 0.150948-00 0.1249511-06 0.124910-01 0.1249501-05 0.150948-00 0.1249501-05 0.150948-00 0.1249501-05 0.150948-00 0.1249501-05 0.150948-00 0.1249501-05 0.150948-00 0.1249501-05 0.150948-00 0.1249501-05 0.150948-00 0.1249501-05 0.150948-00 0.1249501-05 0.150948-00 0.1249501-05 0.150948-00 0.1249501-05 0.
-C.2455CH-C4 0.384CH-C5 0.13739GC-C0 0.1112255-C1 0.134122-C0 0.137224E-C1 0.134122-C0 0.137224E-C1 0.13739GC-C0 0.121255-C1 0.13739G-C0 0.121236-C1 0.13739G-C0 0.121234E-C1 0.13739G-C0 0.121234E-C1 0.13739G-C0 0.13737G-C0
0.121255-C1 0.38122F00 0.12181F00 0.12181F00 0.12181F00 0.121818181818181818181818181818181818181
-0.10776-0.1 -0.10
-0.1310:17-32 0.10:14416-00 0.1354736-00 0.15734746-03 0.755048-04 0.1444596-03 0.755048-04 0.1444596-03 0.755048-04 0.1444596-03 0.755028-04 0.1444596-03 0.755028-04 0.1444596-03 0.755028-03 0.755038-04 0.7550
-0.157314E-C2 C.57505E-D1 C.14445E-C0 C.157314E-C2 C.57531E-C1 C.157234E-C2 C.157314E-C3 C.157331E-C1 C.157231E-C1 C.157231E-C1 C.157231E-C3 C.15723
0.15723416-03 0.1543282-01 0.154238-00 0.15573416-03 0.154238-03 0.15473416-03 0.15473428-01 0.15992828-03 0.15473416-03 0.15473
0.184785 03 0.002055 03 0.184785 00 0.1847875 00 0.184
0.871578E-G4 0.275GGE-D2 0.10885E-D0 0.205178E-G4 0.2952E-D2 0.10885E-D0 0.205178E-D0 0.205178E-
- ************************************
0.20318F-64 0.97249F-93 0.170394F-00 0.18140EF-00 0.141777F-00 0.14177
C.4680754FC95 C.452754FC95 C.452754FC95 C.45404FC95 C.45404FC96 C
0.22.39% 0.3 0.45% 0.4 0.4 0.4 0.4 0.4 0.4 0.4 0.4 0.4 0.4
-04 0.131486-04 0.1484546-00 0.048456-00 0.0
-04 0.338043F-04 0.148854E-00 -04 0.208340F-04 0.167816E-00 -07 0.131514F-04 0.164833F-00
-04 C.ZC#340E-04 D.187816E-00 -57 D.181518E-04 D.1848316-00
-57 0.13151a5-04 0.14a316-00
DO 357788100 FO 354747400
61.0 00-3601691.0 60-36148460 0-3610967.0-

UNCLASSIFIEL

UNCLASSI fied